









LMH1208 SNLS569C - MARCH 2017-REVISED MAY 2020

# LMH1208 12G UHD-SDI Dual Output Cable Driver

#### **Features**

- Supports ST-2082-1 (12G), ST-2081-1 (6G), ST-424 (3G), ST-292 (HD), and ST-259 (SD)
- Compatible with DVB-ASI and AES10 (MADI)
- Dual differential output cable drivers
- On-chip 75-Ω termination and return loss compensation network
- Host-side equalized 100-Ω loop-back output
- Programmable slew rate control on 75-Ω outputs
- Programmable pre-emphasis and output amplitude on 75-Ω outputs
- Programmable de-emphasis and output amplitude on  $100-\Omega$  output
- Polarity inversion on 75- $\Omega$  and 100- $\Omega$  outputs
- Automatic power save in absence of input signal
  - Power consumption: 25 mW (Typical)
- Power-down control through ENABLE pin
- Single 2.5-V supply
  - Power consumption: 200 mW (typical)
- Programmable through pins, SPI, or SMBus interface
- -40°C to +85°C operating temperature range
- 5-mm × 5-mm, 32-pin WQFN package

# **Applications**

- SMPTE compatible serial digital interface
- UHDTV/4K/8K/HDTV/SDTV video
- IP media gateway
- Digital video processing and editing

# Description

LMH1208 device is a 12G UHD-SDI low power dual output cable driver. It supports SMPTE video rates up to 11.88 Gbps, enabling UHD video for 4K/8K applications. The additional equalized  $100-\Omega$  driver output on the host-side can be used for monitoring or signal distribution purposes.

The programmable PCB input equalizer provides high frequency boost to reduce inter symbol interference (ISI) induced by PCB board trace. The integrated 75- $\Omega$  termination and return loss network at both cable driver outputs facilitate overall system design in meeting stringent SMPTE return loss requirements.

Input Signal Detect determines the presence of a valid signal at the input of the cable driver. This sensing feature may be used to alert the user of a system fault and activate a Power Save Mode, reducing the power consumption of the cable driver. The LMH1208 provides selectable slew rates for data rates up to 11.88 Gbps. The output slew rate and amplitude can be controlled by pin, SPI, or SMBus control.

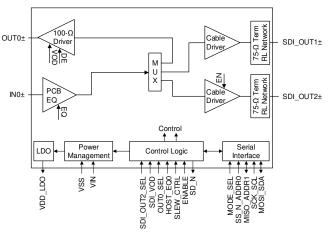
The LMH1208 is powered from a single 2.5-V supply. It is offered in a small footprint 5 mm x 5 mm 32-pin WQFN package. The LMH1208 is pin compatible with the LMH1228 (12G Dual Cable Driver with Integrated Reclocker).

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LMH1208	WQFN (32)	5.00 mm × 5.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### Simplified Block Diagram



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# **Table of Contents**

1	Features 1		7.4 Device Functional Modes	19
2	Applications 1		7.5 Register Maps	23
3	Description 1	8	Application and Implementation	. 24
4	Revision History2		8.1 Application Information	. 24
5	Pin Configuration and Functions3		8.2 Typical Applications	. 25
6	Specifications 6	9 I	Power Supply Recommendations	. 32
Ū	6.1 Absolute Maximum Ratings	10	Layout	. 33
	6.2 ESD Ratings		10.1 Layout Guidelines	. 33
	6.3 Recommended Operating Conditions		10.2 Layout Example	. 35
	6.4 Thermal Information	11	Device and Documentation Support	. 36
	6.5 Electrical Characteristics		11.1 Documentation Support	
	6.6 Recommended SMBus Interface Timing		11.2 Receiving Notification of Documentation Update	s <mark>36</mark>
	Specifications		11.3 Community Resources	. 36
	6.7 Serial Parallel Interface (SPI) Timing		11.4 Trademarks	. 36
	Specifications11		11.5 Electrostatic Discharge Caution	36
	6.8 Typical Characteristics		11.6 Export Control Notice	36
7	Detailed Description 14		11.7 Glossary	. 36
	7.1 Overview 14	12	Mechanical, Packaging, and Orderable	
	7.2 Functional Block Diagram		nformation	. 36
	7.3 Feature Description		12.1 Package Option Addendum	. 37

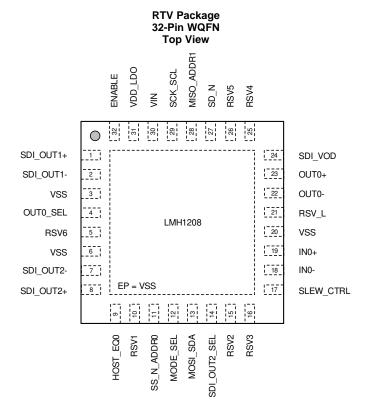
# 4 Revision History

Changes from Revision B (October 2019) to Revision C	Page
Changed HBM Ratings and added pin 27 description	
Changes from Revision A (September 2017) to Revision B	Page
Initial Public Release	

Submit Documentation Feedback



# 5 Pin Configuration and Functions



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## **Pin Functions**

PIN		I/O <sup>(1)</sup>	DESCRIPTION			
NAME NO.		1/0(.)	DESCRIPTION			
HIGH-SPEED DIF	FERENTI	AL I/OS				
SDI_OUT1+	1	I/O, Analog	ingle-ended complementary outputs with on-chip 75-Ω termination at SDI_OUT1+ and			
SDI_OUT1-	2	I/O, Analog	SDI_OUT1 SDI_OUT1± include integrated return loss networks designed to meet the SMPTE output return loss requirements. Connect SDI_OUT1+ to a BNC through a 4.7-μF, AC-coupling capacitor. SDI_OUT1- should be similarly AC-coupled and terminated with an external 4.7-μF capacitor and 75-Ω resistor to GND.			
SDI_OUT2+	8	O, Analog	Single-ended complementary outputs with on-chip 75-Ω termination at SDI_OUT2+ and			
SDI_OUT2-	7	O, Analog	SDI_OUT2 SDI_OUT2± include integrated return loss networks designed to meet the SMPTE output return loss requirements. SDI_OUT2± is used as a second cable driver. Connect SDI_OUT2+ to a BNC through a 4.7-μF, AC-coupling capacitor. SDI_OUT2- should be similarly AC-coupled and terminated with an external 4.7-μF capacitor and 75-Ω resistor to GND.			
IN0+	19	I, Analog	Differential inputs from host video processor. On-chip $100-\Omega$ differential termination.			
INO-	18	I, Analog	Requires external 4.7-µF, AC-coupling capacitors for SMPTE applications.			
OUT0+	23	O, Analog	Differential outputs to host video processor. On-chip 100-Ω differential termination. Requires			
OUT0-	22	O, Analog	external 4.7-µF, AC-coupling capacitors for SMPTE applications.			
CONTROL PINS						
OUT0_SEL	4	I, LVCMOS	OUT0_SEL enables the use of the 100- $\Omega$ host-side output driver at OUT0±. See Table 2 for details. OUT0_SEL is internally pulled high by default (OUT0 disabled).			
HOST_EQ0	9	I, 4-LEVEL	HOST_EQ0 selects the equalizer setting for IN0±. See Table 4 for details.			

(1) I = Input, O = Output, I/O = Input or Output, OD = Open Drain, LVCMOS = 2-State Logic, 4-LEVEL = 4-State Logic



# Pin Functions (continued)

PIN		(4)	,		
NAME	NO.	I/O <sup>(1)</sup>	DESCRIPTION		
MODE_SEL	12	I, 4-LEVEL	MODE_SEL enables the SPI or SMBus serial control interface. See Table 8 for details.		
SDI_OUT2_SEL	14	I, LVCMOS	SDI_OUT2_SEL enables the use of the 75- $\Omega$ output driver at SDI_OUT2±. See Table 2 for details. SDI_OUT2_SEL is internally pulled high by default (SDI_OUT2 disabled).		
SLEW_CTRL	17	I, 4-LEVEL	SLEW_CTRL selects the edge rate for both cable driver outputs. SLEW_CTRL settings are dependent on the operating SMPTE data rate.  SLEW_CTRL also determines the pre-emphasis level applied to both cable driver outputs. See Table 6 and Table 7 for details.		
SDI_VOD	24	I, 4-LEVEL	SDI_VOD selects one of four output amplitudes for the cable drivers at SDI_OUT1± and SDI_OUT2±. See Table 5 for details.		
SD_N	27	O, LVCMOS, OD	SD_N is the Signal Detect indicator. SD_N is pulled low when signal is detected at IN0±. SD_N is a 3.3-V tolerant, open-drain output. It requires an external resistor to a logic supply. SD_N can be reconfigured to indicate Interrupt (INT_N) through register programming. See Status Indicators and Interrupts.		
ENABLE	32	I, LVCMOS	A logic-high at ENABLE enables normal operation for the LMH1208. A logic-low at ENABLE places the LMH1208 in Power-Down Mode.  ENABLE is internally pulled high by default.		
SPI SERIAL CON	ITROL IN	TERFACE, MODE	SEL = F (FLOAT)		
SS_N	11	I, LVCMOS	SS_N is the Slave Select. When SS_N is at logic low, it enables SPI access to the LMH1208 slave device. SS_N is a 2.5-V LVCMOS input and is internally pulled high by default.		
MOSI	13	I, LVCMOS	MOSI is the SPI serial control data input to the LMH1208 slave device when the SPI bus is enabled. MOSI is a 2.5-V LVCMOS input.  An external pullup resistor is recommended.		
MISO	28	O, LVCMOS	MISO is the SPI serial control data output from the LMH1208 slave device. MISO is a 2.5-V LVCMOS output.		
SCK	29	I, LVCMOS	SCK is the SPI serial input clock to the LMH1208 slave device when the SPI interface is enabled. SCK is a 2.5-V LVCMOS input.  An external pullup resistor is recommended.		
SMBUS SERIAL	CONTRO	L INTERFACE, MO	$ODE\_SEL = L (1 K\Omega TO VSS)$		
ADDR0	11	Strap, 4-LEVEL	ADDR[1:0] are 4-level straps, read into the device at power up. They are used to select one of the 16 supported SMBus addresses when SMBus is enabled. See Table 9 for details.		
SDA	13	I/O, LVCMOS, OD	SDA is the SMBus bidirectional data line to or from the LMH1208 slave device when SMBus is enabled. SDA is an open-drain I/O and requires an external pullup resistor to the SMBus termination voltage. SDA is 3.3-V tolerant.		
ADDR1	28	Strap, 4-LEVEL	ADDR[1:0] are 4-level straps, read into the device at power up. They are used to select one of the 16 supported SMBus addresses when SMBus is enabled. See Table 9 for details.		
SCL	29	I/O, LVCMOS, OD	SCL is the SMBus input clock to the LMH1208 slave device when SMBus is enabled. It is driven by a LVCMOS open-drain driver from the SMBus master. SCL requires an external pullup resistor to the SMBus termination voltage. SCL is 3.3-V tolerant.		

Product Folder Links: LMH1208

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# Pin Functions (continued)

PIN		I/O <sup>(1)</sup>	DECODINE		
NAME NO.		1/0(-)	DESCRIPTION		
RESERVED					
RSV1 RSV2 RSV3 RSV4 RSV5 RSV6	10 15 16 25 26 5	_	Reserved pins. Do not connect.		
POWER	*				
VSS	3, 6, 20	I, Ground	Ground reference.		
RSV_L	21	I, Power	Connect RSV_L to the same 2.5-V ± 5% supply as VIN.		
VIN	30	I, Power	VIN is connected to an external 2.5-V ± 5% power supply.		
VDD_LDO	31	O, Power	VDD_LDO is the output of the internal 1.8-V LDO regulator. VDD_LDO output requires an external 1-μF and 0.1-μF bypass capacitor to VSS. The internal LDO is designed to power internal circuitry only.		
EP	_	I, Ground	EP is the exposed pad at the bottom of the RTV package. The exposed pad should be connected to the VSS plane through a 3 x 3 via array.		



# 6 Specifications

## 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	MIN	MAX	UNIT
Supply voltage (VIN, RSV_L)	-0.5	2.75	V
Input voltage for 4-level pins	-0.5	2.75	V
Input/output voltage for 2-level control pins	-0.5	2.75	V
SMBus input/output voltage (SDA, SCL)	-0.5	4	V
SPI input/output voltage (SS_N, MISO, MOSI, and SCK)	-0.5	2.75	V
High-speed input/output voltage (IN0±, SDI_OUT1±, OUT0±, SDI_OUT2±)	-0.5	2.75	V
Input current (IN0±)	-30	30	mA
Operating junction temperature		125	°C
Storage temperature, T <sub>stg</sub>	-65	150	°C

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# 6.2 ESD Ratings

				VALUE	UNIT
	Electrostatic	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	All pins except 13,27, and 29	±6000	
V <sub>(ESD)</sub> discharge		Pins 13, 27, and 29	±5000	V	
		Charged-device model (CDM), per JEDEC specification JESD2	2-C101 <sup>(2)</sup>	±1500	

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. Manufacturing with less than 500-V HBM is possible with the necessary precautions.

#### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
	Supply voltage	VIN, RSV_L to VSS	2.375	2.5	2.625	V
VDD <sub>SMBUS</sub>	SMBus: SDA, SCL open-drain to	ermination voltage	2.375		3.6	V
.,	Source differential launch amplitude	Before 5-inch board trace to IN0±	300		850	mVp-p
V <sub>IN0_LAUNCH</sub>		Before 20-inch board trace to IN0±	650		1000	
T <sub>JUNCTION</sub>	Operating junction temperature				110	°C
T <sub>AMBIENT</sub>	Ambient temperature		-40	25	85	°C
NTps <sub>max</sub>	(1)	50 Hz to 1 MHz, sinusoidal		< 20		>/
	Maximum supply noise (1)	1.1 MHz to 50 MHz, sinusoidal		< 10		mVp-p

(1) The sum of the DC supply voltage and AC supply noise should not exceed the recommended supply voltage range.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Manufacturing with less than 250-V CDM is possible with the necessary precautions.



### 6.4 Thermal Information

		LMH1208	
	THERMAL METRIC <sup>(1)</sup>	RTV (WQFN)	UNIT
		32 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	32.5	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	15.0	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	6.5	°C/W
ΨЈТ	Junction-to-top characterization parameter	0.2	°C/W
ΨЈВ	Junction-to-board characterization parameter	6.5	°C/W
R <sub>0</sub> JC(bot)	Junction-to-case (bottom) thermal resistance	1.7	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

# 6.5 Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER						
		SDI_OUT1± enabled SDI_OUT2± disabled OUT0± disabled		200		mW
PD	Power dissipation Measured with PRBS10,	SDI_OUT1± enabled SDI_OUT2± disabled OUT0± enabled		250		mW
	Operating at 11.88 Gbps, VOD = default	SDI_OUT1± enabled SDI_OUT2± enabled OUT0± disabled		340		mW
		SDI_OUT1± enabled SDI_OUT2± enabled OUT0± enabled		385		mW
$PD_Z$	Power dissipation, Power Save Mode	Power Save Mode, ENABLE = H, no signal applied at IN0±		25		mW
	Current consumption, Measured with PRBS10, Operating at 11.88 Gbps, VOD = default	SDI_OUT1± enabled SDI_OUT2± disabled OUT0± disabled		80	104	mA
		SDI_OUT1± enabled SDI_OUT2± disabled OUT0± enabled		100	125	mA
IDD		SDI_OUT1± enabled SDI_OUT2± enabled OUT0± disabled		136	170	mA
		SDI_OUT1± enabled SDI_OUT2± enabled OUT0± enabled		154	190	mA
IDD <sub>Z</sub>	Current consumption, Power Save Mode	Power Save Mode, ENABLE = H, no signal applied at IN0±		10		mA
IDD <sub>Z_PD</sub>	Current consumption, Power-Down Mode	Power-Down Mode, ENABLE = L, no signal applied at IN0±		10	30	mA
LVCMOS DO	SPECIFICATIONS					
V	Logic high input voltage	2-level input (SS_N, SCK, MOSI, SDI_OUT2_SEL, OUT0_SEL, ENABLE)	0.72 × VIN		VIN + 0.3	V
V <sub>IH</sub>	Logic nigh input voltage	2-level input (SCL, SDA)	0.7 × VIN		3.6	V
$V_{IL}$	Logic low input voltage	2-level input (SS_N, SCK, MOSI, SDI_OUT2_SEL, OUT0_SEL, ENABLE, SCL, SDA)	0		0.3 × VIN	V
V <sub>OH</sub>	Logic high output voltage	IOH = -2 mA, (MISO)	0.8 × VIN		VIN	V



# **Electrical Characteristics (continued)**

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OL}$	Logic low output voltage	IOL = 2 mA, (MISO)	0		0.2 × VIN	V
		IOL = 3 mA, (SD_N, SDA)			0.4	V
		LVCMOS (SDI_OUT2_SEL, ENABLE)			15	μΑ
		LVCMOS (OUT0_SEL)			65	μA
IH	Input high leakage current (Vinput = VIN)	LVCMOS (SD_N)			10	μΑ
	(viiiput = viiv)	SPI mode: LVCMOS (SS_N, SCK, MOSI)			15	μΑ
		SMBus mode: LVCMOS (SCL, SDA)			10	μΑ
		LVCMOS (SDI_OUT2_SEL, ENABLE)	-50			μΑ
		LVCMOS (OUT0_SEL)	-15			μΑ
	Input low leakage current	LVCMOS (SD_N)	-10			μA
IL	(Vinput = GND)	SPI mode: LVCMOS (SCK, MOSI)	-15			μA
		SPI mode: LVCMOS (SS_N)	-50			μA
		SMBus mode: LVCMOS (SCL, SDA)	-10			μA
4-LEVEL LOG	IC DC SPECIFICATIONS (APPLY T	O ALL 4-LEVEL INPUT CONTROL PINS)				
V <sub>LVL_H</sub>	LEVEL-H input voltage	Measured voltage at 4-level pin with external 1 kΩ to VIN		VIN		V
$V_{LVL_{F}}$	LEVEL-F default voltage	Measured voltage 4-level pin at default		2/3 × VIN		V
V <sub>LVL_R</sub>	LEVEL-R input voltage	Measured voltage at 4-level pin with external 20 $k\Omega$ to VSS		1/3 × VIN		V
$V_{LVL_{L}L}$	LEVEL-L input voltage	Measured voltage at 4-level pin with external 1 kΩ to VSS		0		V
Ін	Input high leakage current	4-levels (HOST_EQ0, MODE_SEL, SLEW_CTRL, SDI_VOD)	20	45	80	μΑ
	(Vinput = VIN)	SMBus mode: 4-levels (ADDR0, ADDR1)	20	45	80	μA
l <sub>IL</sub>	Input low leakage current	4-levels (HOST_EQ0, MODE_SEL, SLEW_CTRL, SDI_VOD)	-160	-93	-40	μΑ
	(Vinput = GND)	SMBus mode: 4-levels (ADDR0, ADDR1)	-160	-93	-40	μA
RECEIVER SP	ECIFICATIONS (IN0±)					
R <sub>IN0_TERM</sub>	DC input differential termination	Measured across IN0+ to IN0-	80	100	120	Ω
<del>-</del>		SDD11, 10 MHz – 2.8 GHz		-22		dB
RL <sub>IN0_SDD11</sub>	Input differential return loss (1)	SDD11, 2.8 GHz – 6 GHz		-16		dB
_		SDD11, 6 GHz – 11.1 GHz		-10		dB
RL <sub>IN0_SCD11</sub>	Differential to common-mode input conversion <sup>(1)</sup>	SCD11, 10 MHz to 11.1 GHz		-21		dB
V <sub>IN0_CM</sub>	DC common-mode voltage	Input common-mode voltage at IN0+ or IN0- to GND		2.06		V
CD <sub>ON_IN0</sub>	Signal detect (default) Assert ON threshold level for IN0±	11.88 Gbps, EQ and PLL pathological pattern		20		mVp-p
CD <sub>OFF_IN0</sub>	Signal detect (default) Deassert OFF threshold level for IN0±	11.88 Gbps, EQ and PLL pathological pattern		18		mVp-p
DRIVER OUTP	PUT (SDI_OUT1+ AND SDI_OUT2+)		•			
R <sub>OUT_TERM</sub>	DC output single-ended termination	SDI_OUT1+ and SDI_OUT1-, SDI_OUT2+ and SDI_OUT2- to VIN	63	75	87	Ω

<sup>(1)</sup> This parameter is measured with the LMH1297EVM (Evaluation board for LMH1208).



# **Electrical Characteristics (continued)**

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Output single-ended output	Measure AC signal at SDI_OUT1+ and SDI_OUT2+, with SDI_OUT1- and SDI_OUT2- AC terminated with 75 $\Omega$ SDI_VOD = H		840		mVp-p
$VOD_{CD\_OUTP}$	voltage	SDI VOD = F	720	800	880	mVp-p
		SDI_VOD = R		880		mVp-p
		SDI_VOD = L		760		mVp-p
VOD.	Output single-ended output	Measure AC signal at SDI_OUT1– and SDI_OUT2-, with SDI_OUT1+ and SDI_OUT2+ AC terminated with 75 $\Omega$ SDI_VOD = H		840		mVp-p
VOD <sub>CD_OUTN</sub>	voltage	SDI_VOD = F	720	800	880	mVp-p
		SDI_VOD = R		880		mVp-p
		SDI_VOD = L		760		mVp-p
PRE <sub>CD_OUTP</sub>	Output pre-emphasis	Output pre-emphasis boost amplitude at SDI_OUT1+ and SDI_OUT2+, programmed to maximum setting through register, measured at SDI_VOD = F		2		dB
PRE <sub>CD_OUTN</sub>	Output pre-emphasis	Output pre-emphasis boost amplitude at SDI_OUT1- and SDI_OUT2-, programmed to maximum setting through register, measured at SDI_VOD = F		2		dB
	Output rise and fall time <sup>(1)</sup>	Measured with PRBS10 pattern, default VOD at 20% – 80% amplitude, default preemphasis enabled 11.88 Gbps		34	42	ps
t <sub>R_F_SDI</sub>		5.94 Gbps		36	43	ps
		2.97 Gbps		59	67	ps
		1.485 Gbps		60	73	ps
		270 Mbps	400	550	700	ps
		Measured with PRBS10 pattern, default VOD at 20% – 80% amplitude, default preemphasis enabled 11.88 Gbps		3	18	ps
R_F_DELTA	Output rise and fall time mismatch <sup>(1)</sup>	5.94 Gbps		2.7	12	ps
	monatur	2.97 Gbps		0.8	11	ps
		1.485 Gbps		0.8	12	ps
		270 Mbps		72	150	ps
V <sub>OVERSHOOT</sub>	Output overshoot or undershoot	Measured with PRBS10 pattern, default VOD, default pre-emphasis enabled (2) 12G/6G/3G/HD/SD		5%		
V <sub>DC_OFFSET</sub>	DC offset	12G/6G/3G/HD/SD		±0.2		V
V <sub>DC_WANDER</sub>	DC wander	12G/6G/3G/HD/SD with EQ pathological pattern		20		mV
		S22, 5 MHz to 1.485 GHz		-25		dB
RI on occ	Output return loss at SDI_OUT1+ and SDI_OUT2+	S22, 1.485 GHz to 3 GHz		-22		dB
RL <sub>CD_S22</sub>	reference to 75 $\Omega^{(1)}$	S22, 3 GHz to 6 GHz		-12		dB
		S22, 6 GHz to 12 GHz		-8	T	dB

<sup>(2)</sup> V<sub>OVERSHOOT</sub> overshoot and undershoot maximum measurements are largely affected by the PCB layout and input test pattern. The maximum value specified in *Electrical Characteristics* for V<sub>OVERSHOOT</sub> is based on bench evaluation across temperature and supply voltages with the LMH1297EVM.



# **Electrical Characteristics (continued)**

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DRIVER OUTP	UT (OUT0±)					
R <sub>OUT0_TERM</sub>	DC output differential termination	Measured across OUT0+ and OUT0-	80	100	120	Ω
		Measured with 8T pattern HOST_EQ0 = H		410		mVp-p
VOD <sub>OUT0</sub>	Output differential voltage at OUT0±	HOST_EQ0 = F	485	560	620	mVp-p
	0010±	HOST_EQ0 = R		635		mVp-p
		HOST_EQ0 = L		810		mVp-p
		Measured with 8T pattern HOST_EQ0 = H		410		mVp-p
VOD <sub>OUT0_DE</sub>	De-emphasized output differential voltage at OUT0±	HOST_EQ0 = F	550		mVp-p	
0010_52	differential voltage at OO10±	HOST_EQ0 = R		545		mVp-p
		HOST_EQ0 = L		532		mVp-p
t <sub>R</sub> /t <sub>F</sub>	Output rise and fall time	Measured with 8T Pattern, 20% – 80% amplitude		45		ps
DI.	Output differential return loss <sup>(1)</sup>	Measured with the device powered up and outputs a 10-MHz clock signal SDD22, 10 MHz – 2.8 GHz		-24		dB
RL <sub>OUT0-SDD22</sub>		SDD22, 2.8 GHz – 6 GHz		-16		dB
		SDD22, 6 GHz – 11.1 GHz		-15		dB
RL <sub>OUT0-SCC22</sub>	Output common-mode return loss <sup>(1)</sup>	Measured with the device powered up and outputs a 10-MHz clock signal. SCC22, 10 MHz – 4.75 GHz		-12		dB
		SCC22, 4.75 GHz – 11.1 GHz		-9		dB
V <sub>OUT0_CM</sub>	AC common-mode voltage on OUT0± <sup>(1)</sup>	Default setting, PRBS31, 11.88 Gbps		8		mV (rms)
OUTPUT JITTE	ER					
ADDJ <sub>CD</sub>	Additive jitter <sup>(1)</sup>	Measured at SDI_OUT1+ and SDI_OUT2+, OUT0± disabled PRBS10, 12G/6G/3G/HD/SD		0.03		UI

# 6.6 Recommended SMBus Interface Timing Specifications

over recommended operating supply and temperature ranges unless otherwise specified<sup>(1)</sup>

PARAMETER		TEST CONDITIONS	MIN	NOM MA	UNIT
F <sub>SCL</sub>	SMBUS SCL frequency		10	40	) kHz
T <sub>BUF</sub>	Bus free time between stop and start condition	See Figure 1.	1.3		μs
T <sub>HD:STA</sub>	Hold time after (repeated) start condition	After this period, the first clock is generated.	0.6		μs
T <sub>SU:STA</sub>	Repeated start condition setup time	See Figure 1.	0.6		μs
T <sub>SU:STO</sub>	Stop condition setup time	See Figure 1.	0.6		μs
T <sub>HD:DAT</sub>	Data hold time	See Figure 1.	0		ns
T <sub>SU:DAT</sub>	Data setup time	See Figure 1.	100		ns
T <sub>LOW</sub>	Clock low period	See Figure 1.	1.3		μs
T <sub>HIGH</sub>	Clock high period	See Figure 1.	0.6		μs
T <sub>R</sub>	Clock and data rise time	See Figure 1.		30	) ns
T <sub>F</sub>	Clock and data fall time	See Figure 1.		30	) ns
T <sub>POR</sub>	SMBus ready time after POR	Time from minimum VDDIO to SMBus valid write or read access		5	0 ms

(1) These parameters support SMBus 2.0 specifications.



# 6.7 Serial Parallel Interface (SPI) Timing Specifications

over recommended operating supply and temperature ranges unless otherwise specified<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
F <sub>SCK</sub>	SPI SCK frequency				20	MHz
T <sub>PH</sub>	SCK pulse width high	See Figure 2 and Figure 2	40			% SCK period
T <sub>PL</sub>	SCK pulse width low	See Figure 2 and Figure 3	40			% SCK period
T <sub>SU</sub>	MOSI setup time	See Figure 2 and Figure 2	4			ns
T <sub>H</sub>	MOSI hold time	See Figure 2 and Figure 3	4			ns
T <sub>SSSU</sub>	SS setup time		14			ns
T <sub>SSH</sub>	SS hold time	See Figure 2 and Figure 3	4			ns
T <sub>SSOF</sub>	SS off time		1			μs
$T_{ODZ}$	MISO driven-to-tristate time			20		ns
T <sub>OZD</sub>	MISO tristate-to-driven time	See Figure 2 and Figure 3		10		ns
T <sub>OD</sub>	MISO output delay time			15		ns

### (1) Typical SPI load capacitance is 2 pF.

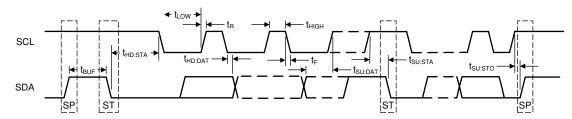


Figure 1. SMBus Timing Parameters

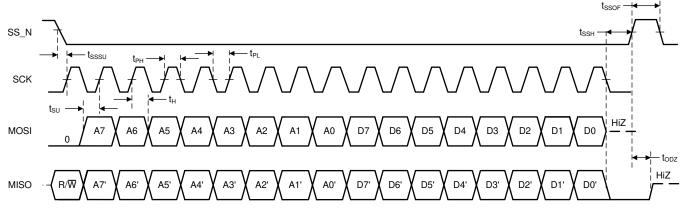


Figure 2. SPI Timing Parameters (Write Operation)

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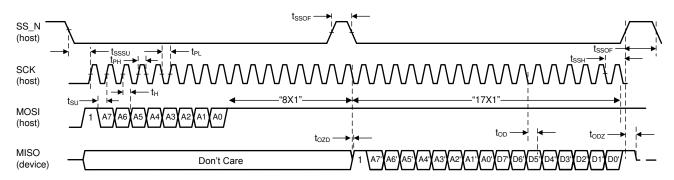
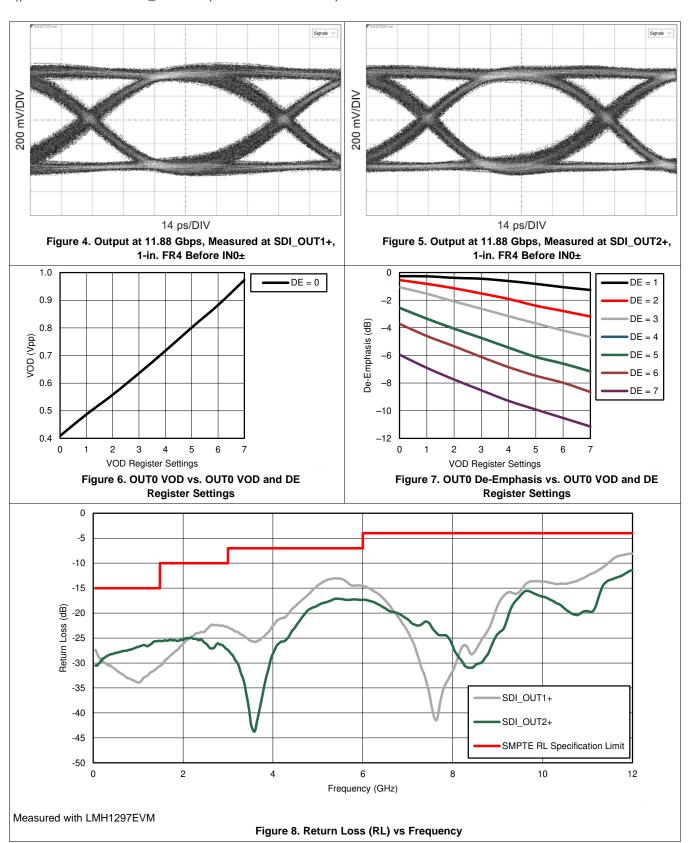


Figure 3. SPI Timing Parameters (Read Operation)



# 6.8 Typical Characteristics

 $T_A = 25^{\circ}C$  and VIN = RSV\_L = 2.5 V (unless otherwise noted)



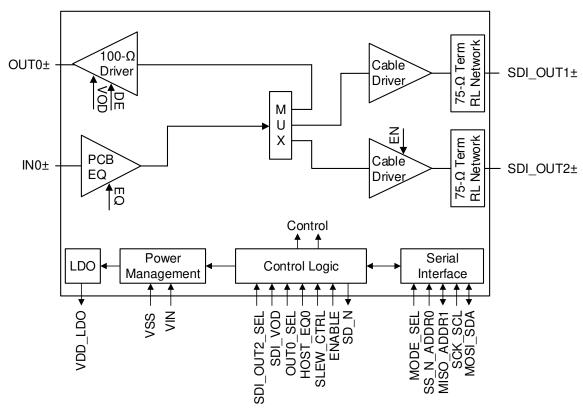


## 7 Detailed Description

#### 7.1 Overview

The LMH1208 is a 12G UHD-SDI dual output cable driver. From the host-side input at IN0 $\pm$ , the signal is equalized and routed to 75- $\Omega$  cable driver outputs at SDI\_OUT1+ and SDI\_OUT2+. The 100- $\Omega$  driver at OUT0 $\pm$  can be used as a host-side loop-back output for monitoring purposes.

## 7.2 Functional Block Diagram



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Figure 9. LMH1208 Block Diagram Overview

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#### 7.3 Feature Description

The LMH1208 data path consists of several key blocks as shown in the functional block diagram. These key blocks are:

- 4-Level Input Pins and Thresholds
- OUT0 SEL and SDI OUT2 SEL Control
- Input Signal Detect
- Continuous Time Linear Equalizer (CTLE)
- Output Driver Control
- Status Indicators and Interrupts

## 7.3.1 4-Level Input Pins and Thresholds

The 4-level input configuration pins use a resistor divider to provide four logic states for each control pin. There is an internal  $30\text{-k}\Omega$  pullup and a  $60\text{-k}\Omega$  pulldown connected to the control pin that sets the default voltage at  $2/3 \times VIN$ . These resistors, together with the external resistor, combine to achieve the desired voltage level. By using the  $1\text{-k}\Omega$  pulldown,  $20\text{-k}\Omega$  pulldown, no connect, and  $1\text{-k}\Omega$  pullup, the optimal voltage levels for each of the four input states are achieved as shown in Table 1.

 LEVEL
 SETTING
 NOMINAL PIN VOLTAGE

 H
 Tie 1 k $\Omega$  to VIN
 VIN

 F
 Float (leave pin open)
 2/3 x VIN

 R
 Tie 20 k $\Omega$  to VSS
 1/3 x VIN

 L
 Tie 1 k $\Omega$  to VSS
 0

Table 1. 4-Level Control Pin Settings

Typical 4-Level Input Thresholds:

- Internal Threshold between L and R = 0.2 x VIN
- Internal Threshold between R and F = 0.5 x VIN
- Internal Threshold between F and H = 0.8 x VIN

### 7.3.2 OUT0\_SEL and SDI\_OUT2\_SEL Control

The OUT0\_SEL and SDI\_OUT2\_SEL pins select the LMH1208 data-path routes. Table 2 shows all possible signal path combinations and typical use cases for each configuration.

Table 2. LMH1208 Signal Path Combinations

OUT0_SEL	SDI_OUT2_SEL	INPUT	MAIN OUTPUT	LINE SIDE SECONDARY OUTPUT	HOST SIDE LOOP-BACK OUTPUT	TYPICAL APPLICATION
Н	Н	IN0±	SDI_OUT1±			Single cable driver
Н	L	IN0±	SDI_OUT1±	SDI_OUT2±		Dual cable drivers
L	Н	IN0±	SDI_OUT1±		OUT0±	Single cable driver with host-side loop-back enabled
L	L	IN0±	SDI_OUT1±	SDI_OUT2±	OUT0±	Dual cable drivers with host-side loop-back enabled

#### 7.3.3 Input Signal Detect

INO has a signal detect circuit to monitor the presence or absence of an input signal. When the input signal amplitude for the selected input exceeds the signal detect assert threshold, the LMH1208 operates in normal operation mode.

In the absence of an input signal, the LMH1208 automatically goes into Power Save Mode to conserve power dissipation. When a valid signal is detected, the LMH1208 automatically exits Power Save Mode and returns to the normal operation mode. If the ENABLE pin is pulled low, the LMH1208 is forced into Power-Down Mode. In Power Save Mode, both the signal detect circuit and the serial interface remain active. In Power-Down Mode, only the serial interface remains active.

Users can monitor the status of the signal detect with the SD\_N pin or through register programming.



**Table 3. Input Signal Detect Modes of Operation** 

ENABLE	SIGNAL INPUT	OPERATING MODE
Н	100-Ω signal input at IN0±	Normal operation Signal Detector at IN0± Serial interface active
Н	No signal at IN0±	Power Save Mode Signal Detector at IN0± Serial interface active
L	Input signal ignored	Power-Down Mode Forced device power down Serial interface active

# 7.3.4 Continuous Time Linear Equalizer (CTLE)

The LMH1208 has a continuous time linear equalizer (CTLE) block for INO. The CTLE compensates for frequency-dependent loss due to the transmission media prior to the device input. The CTLE accomplishes this by applying variable gain to the input signal, thereby boosting higher frequencies more than lower frequencies. The CTLE block extends the signal bandwidth, restores the signal amplitude, and reduces ISI caused by the transmission medium.

IN0 has an on-chip  $100-\Omega$  termination and is designed for AC coupling, requiring a  $4.7-\mu F$ , AC-coupling capacitor for minimizing base-line wander. The PCB equalizer can compensate up to 20 inches of board trace at data rates up to 11.88 Gbps. There is one adapt mode for IN0: AM0 manual mode. In AM0 manual mode, fixed EQ boost settings are applied through user-programmable control.

The HOST\_EQ0 pin determines the IN0 adapt mode and EQ boost level. For normal operation, HOST\_EQ0 = F is recommended. HOST\_EQ0 pin logic settings are shown in Table 4. These HOST\_EQ0 pin settings can be overridden by register control. For more information, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).

Table 4. HOST EQ0 Pin EQ Settings

HOST_EQ0 <sup>(1)</sup>	IN0± EQ BOOST	RECOMMENDED BOARD TRACE IN0± <sup>(2)</sup>
H, F	All Rates: AM0 Manual Mode, EQ=0x00	< 1 inch
R	All Rates: AM0 Manual Mode, EQ=0x80	10-15 inches
L	All Rates: AM0 Manual Mode, EQ=0x90	20 inches

<sup>(1)</sup> The HOST\_EQ0 pin is also used to set OUT0 VOD and de-emphasis values. See Host-Side 100-Ω Output Driver (OUT0±) for more information.

#### 7.3.5 Output Driver Control

### 7.3.5.1 Line-Side Output Cable Driver (SDI\_OUT1+, SDI\_OUT2+)

The LMH1208 has two output cable driver (CD) blocks, one for SDI\_OUT1 and another for SDI\_OUT2. These SDI outputs are designed to drive 75- $\Omega$  single-ended coaxial cables at data rates up to 11.88 Gbps. Both SDI\_OUT1 and SDI\_OUT2 feature an integrated 75- $\Omega$  termination and return loss compensation network for meeting stringent SMPTE return loss requirements (see Figure 8). The cable drivers are designed for AC coupling, requiring a 4.7- $\mu$ F, AC-coupling capacitor for minimizing base-line wander due to the rare-occurring pathological bit pattern.

## 7.3.5.1.1 Output Amplitude (VOD)

SDI\_OUT1 and SDI\_OUT2 are designed for transmission across 75-Ω single-ended impedance. The nominal SDI cable driver output amplitude (VOD) is 800 mVp-p single-ended. In the presence of long output cable lengths or crosstalk, the SDI\_VOD pin can be used to optimize the cable driver output with respect to the nominal amplitude. Table 5 details VOD settings that can be applied to both SDI\_OUT1 and SDI\_OUT2. The SDI\_VOD pin can be overridden through register control. In addition, the nominal VOD amplitude can be changed by register control. For more information, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).

<sup>(2)</sup> Recommended board trace at 11.88 Gbps.



Table 5. SDI\_VOD Settings for Line-Side Output Amplitude

SDI_VOD	DESCRIPTION
Н	about +5% of nominal
F	800 mVp-p (nominal)
R	about +10% of nominal
L	about -5% of nominal

#### 7.3.5.1.2 Output Pre-Emphasis

In addition to SDI cable driver VOD control, the LMH1208 can add pre-emphasis on the cable driver output to improve output signal integrity at a UHD (12G, 6G) or HD (3G, 1.5G) input data rate. By default, pre-emphasis is enabled based on the SLEW CTRL pin setting shown in Table 6.

Table 6. SLEW\_CTRL Settings for SDI\_OUT1 and SDI\_OUT2 Pre-Emphasis

SLEW_CTRL	OUTPUT PRE-EMPHASIS	DATA RATE
Н	Enabled	UHD: 11.88 Gbps UHD: 5.94 Gbps
F	Disabled	HD: 2.97 Gbps HD: 1.485 Gbps
R, L	Disabled	SD: 270 Mbps

When enabled, the amount of pre-emphasis applied to cable driver outputs is determined by register control. When disabled, no pre-emphasis is applied. Pre-emphasis control can be overridden for select or all data rates through register control. For more information, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).

#### 7.3.5.1.3 Output Slew Rate

SMPTE specifications require different output driver rise and fall times depending on the operating data rate. To meet these requirements, the output edge rate of SDI\_OUT1 and SDI\_OUT2 can be configured by the SLEW\_CTRL pin. Table 7 shows the recommended SLEW\_CTRL pin logic setting and typical edge rate at the cable driver output for each data rate.

Table 7. SLEW CTRL Settings for SDI OUT1 and SDI OUT2 Output Edge Rate

SLEW_CTRL DATA RATE		CABLE DRIVER OUTPUT EDGE RATE (TYP)
11	11.88 Gbps	34 ps
Н	5.94 Gbps	36 ps
F	2.97 Gbps	59 ps
F	1.485 Gbps	60 ps
R, L 270 Mbps		550 ps

Users can also program the desired edge rate manually through register control. For more information, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).

### 7.3.5.1.4 Output Polarity Inversion

Polarity inversion is supported on both SDI\_OUT1 and SDI\_OUT2 outputs through register control.

### 7.3.5.2 Host-Side 100-Ω Output Driver (OUT0±)

OUT0 is a  $100-\Omega$  driver output. OUT0 serves as a host-side loop-back output. OUT0 also supports polarity inversion.

The driver offers users the capability to select higher output amplitude and de-emphasis levels for longer board trace that connects the drivers to their downstream receivers. Driver de-emphasis provides transmitter equalization to reduce the ISI caused by the board trace.



The VOD and de-emphasis levels for OUT0 are set by default to 570 mVp-p and -0.4 dB, and these values are recommended for driving 1-2 inches of board trace from OUT0± at 11.88 Gbps. These settings can be changed through register control if desired. When these parameters are controlled by registers, the VOD and de-emphasis levels can be programmed independently. For more information, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).

#### 7.3.6 Status Indicators and Interrupts

The SD\_N pin is a 3.3-V tolerant, active-low, open-drain output. An external resistor to the logic supply is required. The SD\_N pin can be configured to indicate input signal detect or an interrupt event.

#### 7.3.6.1 SD\_N (Signal Detect)

By default, SD\_N indicates a SD\_N (signal detect) event, and this pin asserts low after a valid signal is detected by the IN0 signal detect circuit. For more information about how to reconfigure the SD\_N pin functionality, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).

#### 7.3.6.2 INT N (Interrupt)

The SD\_N pin can be configured to indicate an INT\_N (interrupt) event. When configured as an INT\_N output, the pin asserts low when an interrupt occurs, according to the programmed interrupt masks. Two separate masks can be programmed through register control as interrupt sources:

• If there is a loss of signal (LOS) event on IN0 (2 separate masks).

INT\_N is a sticky bit, meaning that it will flag after an interrupt occurs and will not clear until read-back. Once the Interrupt Status Register is read, the INT\_N pin will assert high again. For more information about how to configure the SD\_N pin for INT\_N functionality, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).



#### 7.4 Device Functional Modes

The LMH1208 operates in one of two modes: System Management Bus (SMBus) or Serial Peripheral Interface (SPI) mode. To determine the mode of operation, the proper setting must be applied to the MODE\_SEL pin at power up, as detailed in Table 8.

Table 8. MODE\_SEL Pin Settings

LEVEL	DESCRIPTION
Н	Reserved for factory testing – do not use
F	Selects SPI Interface for register access
R	Reserved for factory testing – do not use
L	Selects SMBus Interface for register access

### 7.4.1 System Management Bus (SMBus) Mode

The SMBus interface can also be used to control the device. If MODE\_SEL = Low (1 k $\Omega$  to VSS), Pins 13 and 29 are configured as SDA and SCL. Pins 11 and 28 are address straps ADDR0 and ADDR1 during power up. The maximum operating speed supported on the SMBUS pins is 400 kHz.

Table 9. SMBus Device Slave Addresses<sup>(1)</sup>

ADDR0 (LEVEL)	ADDR1 (LEVEL)	7-BIT SLAVE ADDRESS [HEX]	8-BIT WRITE COMMAND [HEX]
L	L	2D	5A
L	R	2E	5C
L	F	2F	5E
L	Н	30	60
R	L	31	62
R	R	32	64
R	F	33	66
R	Н	34	68
F	L	35	6A
F	R	36	6C
F	F	37	6E
F	Н	38	70
Н	L	39	72
Н	R	3A	74
Н	F	3B	76
Н	Н	3C	78

<sup>(1)</sup> The 8-bit write command consists of the 7-bit slave address (Bits 7:1) with 0 appended to the LSB to indicate an SMBus write. For example, if the 7-bit slave address is 0x2D (010 1101'b), the 8-bit write command is 0x5A (0101 1010'b).

#### 7.4.1.1 SMBus Read and Write Transaction

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SMBus is a two-wire serial interface through which various system component chips can communicate with the master. Slave devices are identified by having a unique device address. The two-wire serial interface consists of SCL and SDA signals. SCL is a clock output from the master to all of the slave devices on the bus. SDA is a bidirectional data signal between the master and slave devices. The LMH1208 SMBus SCL and SDA signals are open-drain and require external pullup resistors.

#### Start and Stop:

The master generates Start and Stop patterns at the beginning and end of each transaction.

- Start: High-to-low transition (falling edge) of SDA while SCL is high.
- Start: High-to-low transition (falling edge) of SDA while SCL is high.

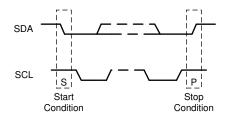


Figure 10. Start and Stop Conditions

The master generates nine clock pulses for each byte transfer. The 9th clock pulse constitutes the ACK cycle. The transmitter releases SDA to allow the receiver to send the ACK signal. An ACK is recorded when the device pulls SDA low, while a NACK is recorded if the line remains high.

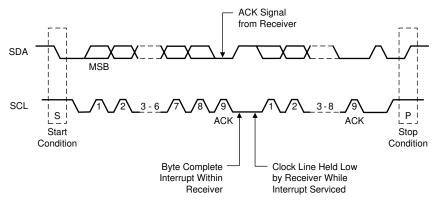


Figure 11. Acknowledge (ACK)

#### 7.4.1.1.1 SMBus Write Operation Format

Writing data to a slave device consists of three parts, as illustrated in Figure 12:

- 1. The master begins with a start condition followed by the slave device address with the  $R/\overline{W}$  bit set to 0'b.
- 2. After an ACK from the slave device, the 8-bit register word address is written.
- 3. After an ACK from the slave device, the 8-bit data is written, followed by a stop condition.

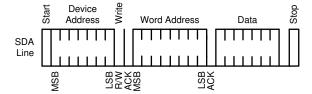


Figure 12. SMBus Write Operation

#### 7.4.1.1.2 SMBus Read Operation Format

SMBus read operation consists of four parts, as illustrated in Figure 13:

- 1. The master begins with a start condition, followed by the slave device address with the  $R/\overline{W}$  bit set to 0'b.
- 2. After an ACK from the slave device, the 8-bit register word address is written.
- 3. After an ACK from the slave device, the master initiates a restart condition, followed by the slave address with the R/W bit set to 1'b.
- 4. After an ACK from the slave device, the 8-bit data is read-back. The last ACK is high if there are no more bytes to read, and the last read is followed by a stop condition.

Product Folder Links: LMH1208

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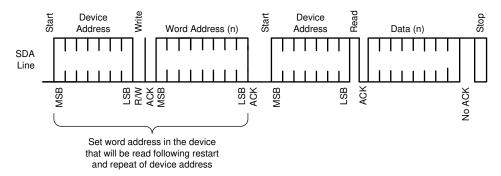


Figure 13. SMBus Read Operation

## 7.4.2 Serial Peripheral Interface (SPI) Mode

If MODE\_SEL = F or H, the LMH1208 is in SPI mode. In SPI mode, the following pins are used for SPI bus communication:

- MOSI (Pin 13): Master Output Slave Input
- MISO (Pin 28): Master Input Slave Output
- SS\_N (Pin 11): Slave Select (Active Low)
- SCK (Pin 29): Serial Clock (Input to the LMH1208 Slave Device)

#### 7.4.2.1 SPI Read and Write Transactions

Each SPI transaction to a single device is 17 bits long and is framed by SS\_N when asserted low. The MOSI input is ignored, and the MISO output is floated whenever SS\_N is deasserted (high).

The bits are shifted in left-to-right. The first bit is R/W, which is 1'b for *read* and 0'b for *write*. Bits A7-A0 are the 8-bit register address, and bits D7-D0 are the 8-bit read or write data. The previous SPI command, address, and data are shifted out on MISO as the current command, address, and data are shifted in on MOSI. In all SPI transactions, the MISO output signal is enabled asynchronously when SS\_N asserts low. The contents of a single MOSI or MISO transaction frame are shown in Figure 14.

Figure 14. 17-Bit Single SPI Transaction Frame

R/W	A7	A6	A5	A4	А3	A2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0	
-----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	--



### 7.4.2.2 SPI Write Transaction Format

For SPI writes, the R/W bit is 0'b. SPI write transactions are 17 bits per device, and the command is executed on the rising edge of SS\_N. The SPI transaction always starts on the rising edge of the clock.

The signal timing for a SPI Write transaction is shown in Figure 2. The *prime* values on MISO (for example, A7') reflect the contents of the shift register from the previous SPI transaction and are listed as don't care for the current transaction.

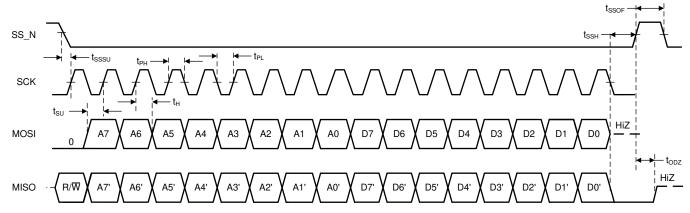


Figure 15. Signal Timing for a SPI Write Transaction

#### 7.4.2.3 SPI Read Transaction Format

A SPI read transaction is 34 bits per device and consists of two 17-bit frames. The first 17-bit read transaction frame shifts in the address to be read, followed by a dummy transaction second frame to shift out 17-bit read data. The R/W bit is 1'b for the read transaction, as shown in Figure 3.

The first 17 bits from the read transaction specifies 1-bit of R/W and 8-bits of address A7-A0 in the first 8 bits. The eight 1's following the address are ignored. The second dummy transaction acts like a read operation on address 0xFF and needs to be ignored. However, the transaction is necessary in order to shift out the read data D7-D0 in the last 8 bits of the MISO output. As with the SPI Write, the *prime* values on MISO during the first 16 clocks are listed as don't care for this portion of the transaction. The values shifted out on MISO during the last 17 clocks reflect the read address and 8-bit read data for the current transaction.

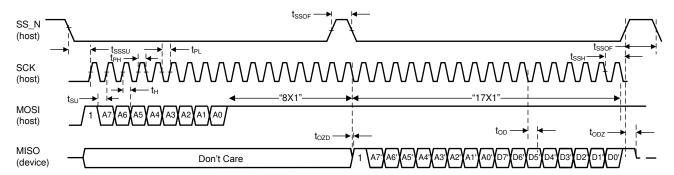


Figure 16. Signal Timing for a SPI Read Transaction



### 7.4.2.4 SPI Daisy Chain

The LMH1208 supports SPI daisy-chaining among multiple devices, as shown in Figure 17.

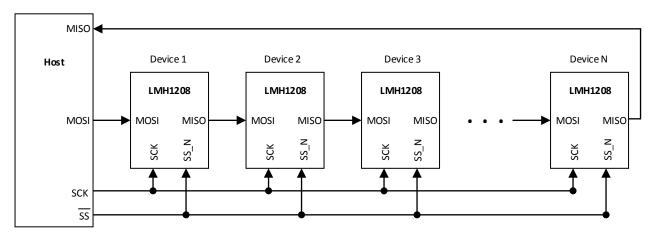


Figure 17. Daisy-Chain Configuration

Each LMH1208 device is directly connected to the SCK and SS\_N pins of the host. The first LMH1208 device in the chain is connected to the host's MOSI pin, and the last device in the chain is connected to the host's MISO pin. The MOSI pin of each intermediate LMH1208 device in the chain is connected to the MISO pin of the previous LMH1208 device, thereby creating a serial shift register. In a daisy-chain configuration of N  $\times$  LMH1208 devices, the host conceptually sees a shift register of length 17  $\times$  N for a basic SPI transaction, during which SS N is asserted low for 17  $\times$  N clock cycles.

## 7.5 Register Maps

The LMH1208 register map is divided into three register pages. These register pages are used to control different aspects of the LMH1208 functionality. A brief summary of the pages is shown below:

- 1. **Share Register Page:** This page corresponds to global parameters, such as LMH1208 device ID and SD\_N status configuration. This is the default page at start-up. Access this page by setting Reg 0xFF[2:0] = 000'b.
- 2. **CTLE Register Page:** This page corresponds to IN0 PCB CTLE, output mux settings, and output interrupt overrides. Access this page by setting Reg 0xFF[2:0] = 100'b.
- 3. **Driver Register Page:** This page corresponds to OUT0, SDI\_OUT1, and SDI\_OUT2 driver output settings. Access this page by setting Reg 0xFF[2:0] = 101'b.

For the complete register map, typical device configurations, and proper register reset sequencing, refer to the *LMH1228 and LMH1208 Programming Guide* (SNAU206).



# 8 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

## 8.1 Application Information

## 8.1.1 SMPTE Requirements and Specifications

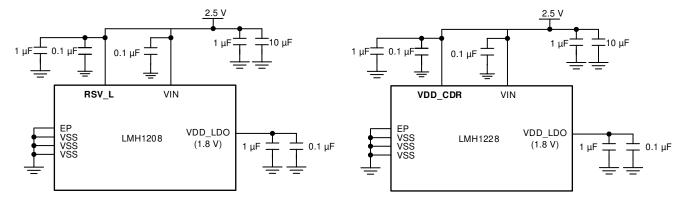
SMPTE specifies several key requirements for the Serial Digital Interface to transport digital video over coaxial cables. Such requirements include return loss, AC coupling, and data rate dependency with rise and fall times.

- 1. **Return Loss:** This specification details how closely the port resembles 75-Ω impedance across a specified frequency band. The LMH1208 features a built-in 75-Ω return-loss network on SDI\_OUT1 and SDI\_OUT2 to minimize parasitics and improve overall signal integrity.
- 2. **AC Coupling:** AC-coupling capacitors are required for transporting uncompressed serial data streams with heavy low-frequency content. The use of 4.7- $\mu$ F, AC-coupling capacitors is recommended to avoid low-frequency DC wander.
- 3. Rise/Fall Time: Output  $75-\Omega$  signals are required to meet certain rise and fall timing depending on the data rate. This improves the eye opening observed for the receiving device. The LMH1208 SDI\_OUT1 and SDI\_OUT2 cable drivers feature programmable edge rate adjustment to meet SMPTE rise and fall time requirements.

TI recommends placing the LMH1208 as close as possible to the 75- $\Omega$  BNC ports to meet SMPTE specifications.

## 8.1.2 LMH1208 and LMH1228 Compatibility

The LMH1208 is pin-compatible with the LMH1228 (12G UHD-SDI Dual Output Cable Driver with Integrated Reclocker) when LMH1208 RSV\_L pin is tied to 2.5 V. This pin compatibility allows users to upgrade easily to the LMH1228. See Figure 18 for details.



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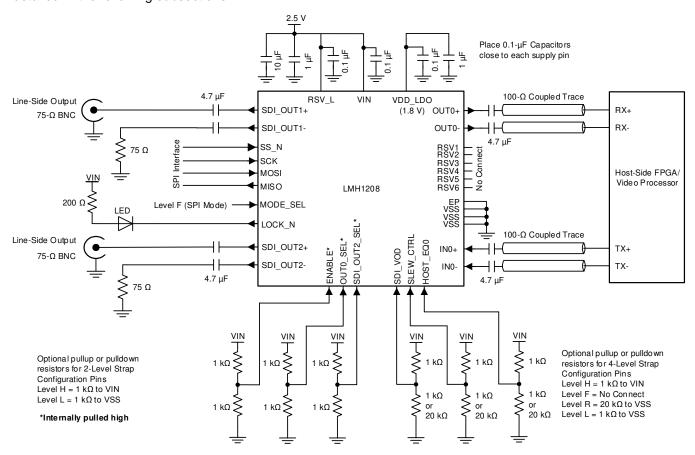
Figure 18. Pin Connections for LMH1208 and LMH1228 Compatibility



## 8.2 Typical Applications

The LMH1208 is a dual cable driver that supports SDI data rates up to 11.88 Gbps. Figure 19 shows a typical application circuit for the LMH1208.

Specific examples of typical applications for the LMH1208 as a dual cable driver and distribution amplifier are detailed in the following subsections.



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Figure 19. LMH1208 Typical Application Circuit

Product Folder Links: LMH1208

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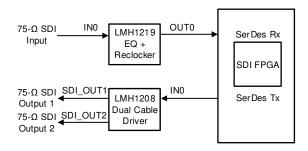


## **Typical Applications (continued)**

#### 8.2.1 Dual Cable Driver

The LMH1208 can be configured as a dual cable driver to route the same SDI output signal to multiple receivers. In this configuration, the LMH1208 is programmed to equalize 100-Ω SDI input data at INO and uses the dual cable drivers at SDI\_OUT1 and SDI\_OUT2 to drive out the SDI signal.

Figure 20 shows a typical application of an LMH1208 as a dual cable driver output. In this example, the LMH1219 Cable EQ with Integrated Reclocker provides an SDI input to the SDI FPGA. The FPGA then sends post-processed SDI data to the IN0 of the LMH1208, which drives the data on cable driver outputs SDI\_OUT1 and SDI\_OUT2.



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Figure 20. LMH1208 Dual Cable Driver Application

#### 8.2.1.1 Design Requirements

For general LMH1208 design requirements, reference the guidelines in Table 10.

For dual cable driver application-specific requirements, reference the guidelines in Table 11.

**DESIGN PARAMETER** REQUIREMENTS SDI\_OUT1+, SDI\_OUT2+ AC-coupling capacitors 4.7- $\mu F$  capacitors recommended SDI\_OUT1-, SDI\_OUT2- AC-coupling capacitors 4.7-μF capacitors recommended, AC terminated with 75  $\Omega$  to VSS. IN0± and OUT0± AC-coupling capacitors 4.7-μF capacitors recommended Input and output terminations Input and output terminations provided internally. Do not add external terminations. 10-μF and 1-μF bulk capacitors; place close to each device. DC power supply decoupling capacitors 0.1-μF capacitor; place close to each supply pin. VDD\_LDO decoupling capacitors  $1-\mu F$  and  $0.1-\mu F$  capacitors; place as close as possible to the device VDD\_LDO pin. SPI: Leave MODE\_SEL unconnected (Level F) MODE\_SEL pin SMBus: Connect 1 kΩ to VSS (Level L)

Table 10. LMH1208 General Design Requirements

Table 11. LMH1208 Dual Cable Driver Requirements

DESIGN PARAMETER	REQUIREMENTS
OUT0_SEL pin	1 k $\Omega$ to VIN (Level H) to disable the OUT0 loop-back output
SDI_OUT2_SEL pin	1 k $\Omega$ to VSS (Level L) to enable SDI_OUT2 as secondary cable output

#### 8.2.1.2 Detailed Design Procedure

The design procedure for dual cable driver applications is as follows:

- 1. Select a power supply that meets the DC and AC requirements in Recommended Operating Conditions.
- Choose a small 0402 surface mount ceramic capacitor for AC-coupling capacitors to maintain characteristic impedance.
- 3. Choose a high-quality,  $75-\Omega$  BNC connector that is capable of supporting 11.88-Gbps applications. Consult a BNC supplier regarding insertion loss, impedance specifications, and recommended footprint for meeting



SMPTE return loss.

- 4. Follow detailed high-speed layout recommendations provided in *Layout Guidelines* to ensure optimal signal quality when interconnecting 75- $\Omega$  and 100- $\Omega$  signals to the LMH1208.
- 5. Determine whether SPI or SMBus communication is necessary. If the LMH1208 must be programmed with settings other than what is offered by pin control, users must use SPI or SMBus Mode for additional programming.
- 6. Configure OUT0\_SEL and SDI\_OUT2\_SEL pins according to the desired default use case.
- 7. Tune the SDI\_VOD output amplitude control pin for optimal signal quality depending on the cable length attached at SDI\_OUT1+ and SDI\_OUT2+. Use register control for more tuning options, if necessary.

#### 8.2.1.3 Application Curves

The LMH1208 performance on SDI\_OUT1+ and SDI\_OUT2+ was measured with the test setups shown in Figure 21 and Figure 22.

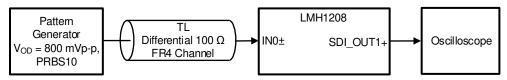


Figure 21. Test Setup for LMH1208 to SDI\_OUT1+

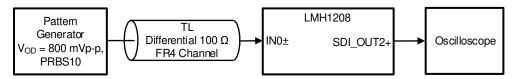
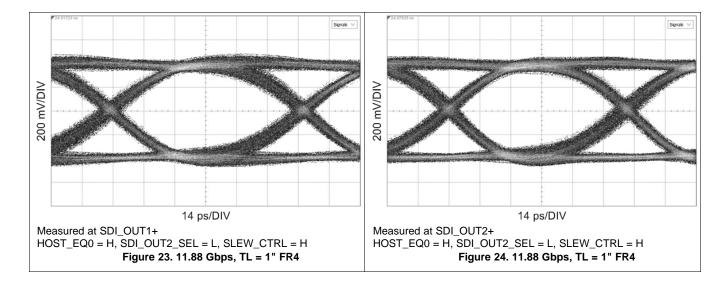
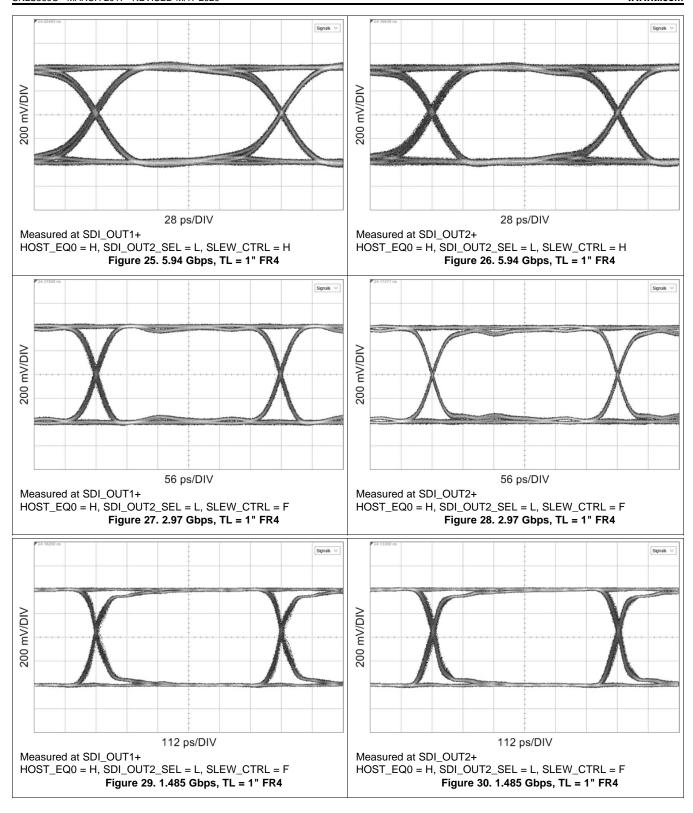


Figure 22. Test Setup for LMH1208 to SDI OUT2+

The eye diagrams in this subsection show how the LMH1208 improves overall signal integrity in the data path for  $100-\Omega$  differential FR4 PCB trace at IN0±.



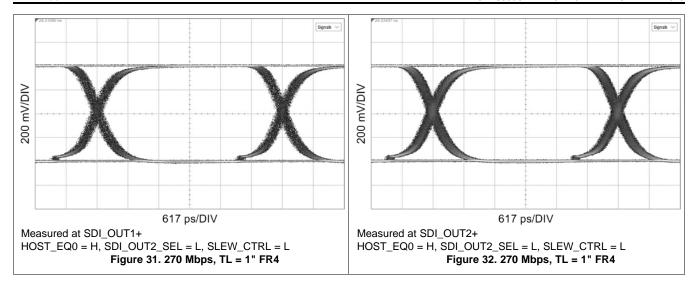




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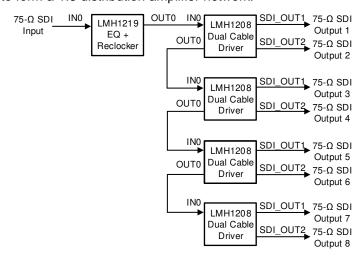




#### 8.2.2 Distribution Amplifier

The LMH1208 can be configured as a distribution amplifier to distribute the same SDI input signal to multiple cable driver outputs. In this configuration, the LMH1208 uses the dual cable drivers at SDI\_OUT1 and SDI\_OUT2 to drive out the SDI signal seen at IN0. Meanwhile, the loop-back output on OUT0 is daisy-chained as a duplicate input to IN0 of the next LMH1208.

Figure 33 shows a typical application where four LMH1208s are used in combination with an LMH1219 Cable EQ with Integrated Reclocker to form a 1:8 distribution amplifier network.



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Figure 33. LMH1208 Distribution Amplifier Application

#### 8.2.2.1 Design Requirements

See Table 10 in Dual Cable Driver Design Requirements for general LMH1208 design requirements.

For distribution amplifier application-specific requirements, reference the guidelines in Table 12.

Table 12. LMH1208 Distribution Amplifier Requirements

DESIGN PARAMETER	REQUIREMENTS
OUT0_SEL pin	1 k $\Omega$ to VSS (Level L) to enable OUT0 as a loop-back output to the next LMH1208 IN0 input
SDI_OUT2_SEL pin	1 kΩ to VSS (Level L) to enable SDI_OUT2 as secondary cable output



### 8.2.2.2 Detailed Design Procedure

See *Dual Cable Driver Detailed Design Procedure* and follow Steps 1 through 5. Refer to the additional steps below for distribution amplifier applications.

- 1. Configure OUT0\_SEL and SDI\_OUT2\_SEL pins according to the desired default use case.
- 2. Tune the output VOD and de-emphasis level for the 100-Ω driver prior to each LMH1208 IN0±. In the distribution amplifier example shown in Figure 33, this step applies to the LMH1219 OUT0± driver and all LMH1208 OUT0± drivers that are daisy-chained to a subsequent LMH1208 IN0±. If OUT0± is located within 1-2 inches of IN0±, then use a lower VOD setting and no de-emphasis. If the OUT0± is located many inches away from IN0±, some VOD gain and de-emphasis may be required at OUT0± for the IN0 CTLE to equalize optimally. Use register control for more tuning options if necessary.
- 3. Tune the SDI\_VOD output amplitude control pin for optimal signal quality depending on the cable length attached at SDI\_OUT1+ and SDI\_OUT2+ for each LMH1208. Use register control for more tuning options, if necessary.

#### 8.2.2.3 Application Curves

The LMH1208 performance on OUT0± was measured with the test setup shown in Figure 21.

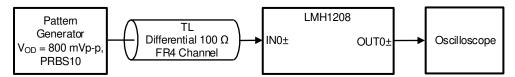
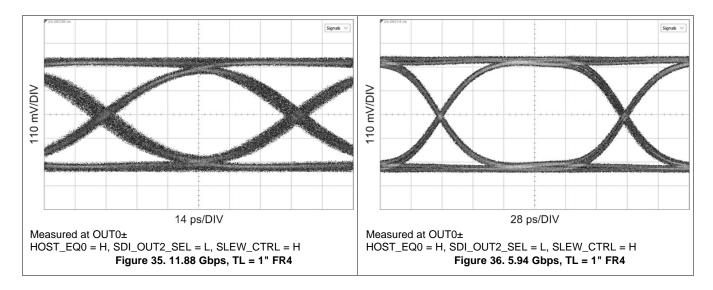
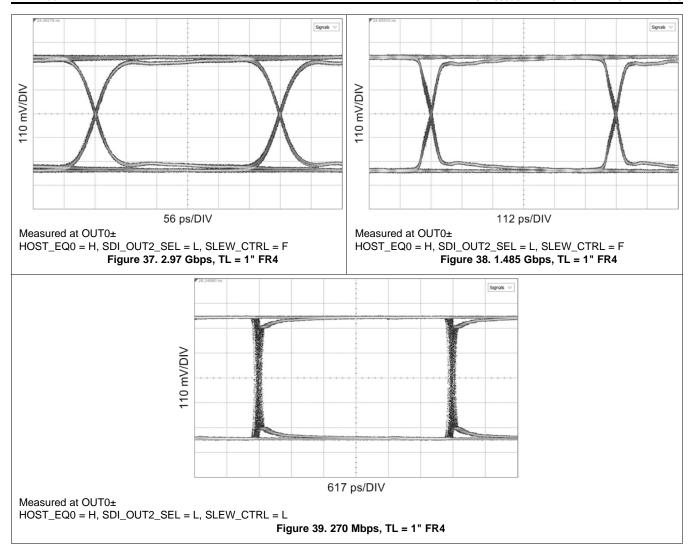


Figure 34. Test Setup for LMH1208 to OUT0±

The eye diagrams in this subsection show how the LMH1208 improves overall signal integrity in the data path for  $100-\Omega$  differential FR4 PCB trace at IN0±.





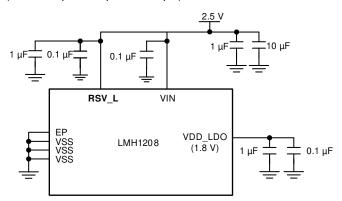


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# 9 Power Supply Recommendations

The LMH1208 requires decoupling capacitors to ensure a stable power supply. For power supply decoupling, 0.1- $\mu$ F surface-mount ceramic capacitors must be placed close to each RSV\_L, VDD\_LDO and VIN supply pin to VSS. Larger bulk capacitors (for example, 10  $\mu$ F and 1  $\mu$ F) are recommended for RSV\_L and VIN.



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Figure 40. Recommended Power Supply Decoupling

Good supply bypassing requires low inductance capacitors. This can be achieved through an array of multiple small body size surface-mount bypass capacitors to keep low supply impedance. Better results can be achieved through the use of a buried capacitor formed by a VDD and VSS plane separated by 2 to 4 mil dielectric in a printed-circuit board.



# 10 Layout

### 10.1 Layout Guidelines

The following guidelines are recommended to optimize the board layout for the LMH1208.

#### 10.1.1 Board Stack-Up and Ground References

- Choose a suitable board stack-up that supports 75- $\Omega$  single-ended trace and 100- $\Omega$  differential trace routing on the top layer of the board. This is typically done with a Layer-2 ground plane reference for the 100- $\Omega$  differential traces and a Layer-3 ground plane reference for the 75- $\Omega$  single-end traces.
- Maintain a distance of at least 5 times the trace width between signal trace and ground reference if they are
  on the same layer. This prevents unwanted changes in the characteristic impedance.
- Maintain a consistent ground plane reference for each high-speed trace from source to end-point. Ground reference discontinuities lead to characteristic impedance mismatch.

## 10.1.2 High-Speed PCB Trace Routing and Coupling

Observe the following general high-speed recommendations for high-speed trace routing:

- For differential pairs, maintain a uniform width and gap for each differential pair where possible. When traces
  must diverge (for example, due to AC-coupling capacitors), ensure that the traces branch out or merge
  uniformly.
- To prevent reflections due to trace routing, ensure that trace bends are at most 45°. Right angle bends should be implemented with at least two 45° corners. Radial bends are ideal.
- Avoid using signal vias. If signal vias must be used, a return path (GND) via must be placed near the signal
  via to provide a consistent ground reference and minimize impedance discontinuities.
- Avoid via stubs by back-drilling as necessary.

## SDI\_OUT1± and SDI OUT2±:

- Use an uncoupled trace with 75-Ω single-ended impedance for signal routing to SDI\_OUT1± and SDI\_OUT2±.
- The trace width is typically 8 to 10 mils with reference to a Layer-3 ground plane.

### IN0± and OUT0±:

- Use coupled traces with 100-Ω differential impedance for signal routing to IN0± and OUT0±.
- The trace width is typically 5 to 8 mils with reference to a Layer-2 ground plane.

#### 10.1.3 Anti-Pads

 Place anti-pads (ground relief) on the power and ground planes directly under the 4.7-μF, AC-coupling capacitor and IC landing pads to minimize parasitic capacitance. The size of the anti-pad and the number of layers to use the anti-pad depend on the board stack-up and can be determined by a 3-dimension electromagnetic simulation tool.

### 10.1.4 BNC Connector Layout and Routing

- Use a well-designed BNC footprint to ensure the BNC's signal landing pad achieves 75-Ω characteristic impedance. BNC suppliers usually provide recommendations on BNC footprint for best results.
- Keep trace length short between the BNC and SDI\_OUT1±. The trace routing for SDI\_OUT1+ and SDI\_OUT1- should be as symmetrical as possible, with approximately equal lengths and equal loading. The same is true for SDI\_OUT2+ and SDI\_OUT2-.

# 10.1.5 Power Supply and Ground Connections

- Connect each supply pin (RSV\_L, VIN, VDD\_LDO) directly to the power or ground planes with a short via. The via is usually placed tangent to the supply pins' landing pads with the shortest trace possible.
- Power supply decoupling capacitors should be a small physical size (0402 or smaller) and placed close to the supply pins to minimize inductance. The capacitors are commonly placed on the bottom layer and share the ground of the EP (Exposed Pad).



### **Layout Guidelines (continued)**

#### 10.1.6 Footprint Recommendations

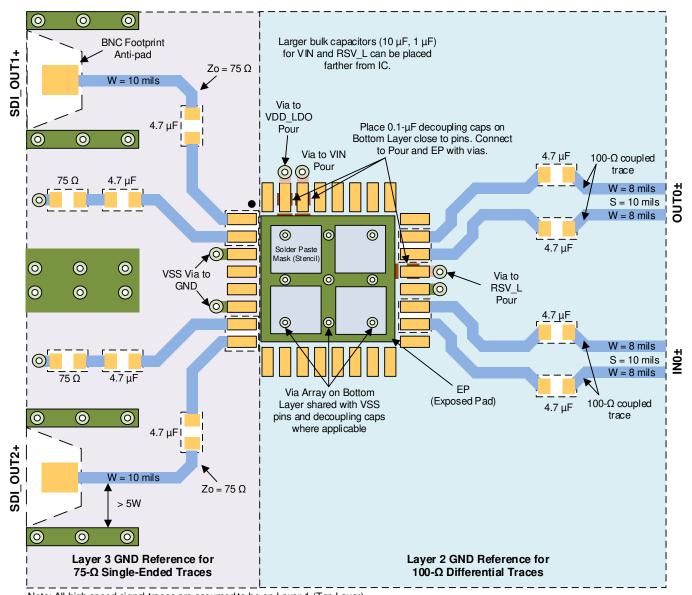
- Stencil parameters for the EP (Exposed Pad) such as aperture area ratio and the fabrication process have a significant impact on paste deposition. Inspection of the stencil prior to placement of the WQFN package is highly recommended to improve board assembly yields. If the via and aperture openings are not carefully monitored, the solder may flow unevenly through the EP. Stencil parameters for aperture opening and via locations are shown in the RTV package drawing in *Mechanical, Packaging, and Orderable Information*.
- The EP of the package must be connected to the ground plane through a 3 x 3 via array. These vias are solder-masked to avoid solder flowing into the plated-through holes during the board manufacturing process. Details about via dimensions are also shown in the RTV package drawing in *Mechanical, Packaging, and Orderable Information*.

More information on the WQFN style package is provided in *QFN/SON PCB Attachment Application Report* (SLUA271).



## 10.2 Layout Example

The example shown in Figure 41 demonstrates the LMH1208 layout guidelines highlighted in Layout Guidelines.



Note: All high speed signal traces are assumed to be on Layer 1 (Top Layer).

Figure 41. LMH1208 High-Speed Trace Layout Example

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# 11 Device and Documentation Support

## 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation, see the following:

QFN/SON PCB Attachment Application Report (SLUA271)

### 11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

## 11.3 Community Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 11.4 Trademarks

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### 11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 11.6 Export Control Notice

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#### 11.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

www.ti.com

# 12.1 Package Option Addendum

#### 12.1.1 Packaging Information

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish <sup>(3)</sup>	MSL Peak Temp (4)	Op Temp (°C)	Device Marking <sup>(5)(6)</sup>
LMH1208RTVR	ACTIVE	WQFN	RTV	32	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	L1208
LMH1208RTVT	ACTIVE	WQFN	RTV	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	L1208

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PRE\_PROD Unannounced device, not in production, not available for mass market, nor on the web, samples not available.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

- (3) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.
- (4) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (5) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device
- (6) Multiple Device markings will be inside parentheses. Only on Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

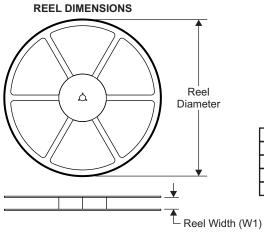
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Product Folder Links: LMH1208



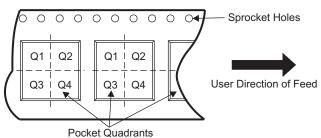
# 12.1.2 Tape and Reel Information



# 

A0	Dimension designed to accommodate the component width
В0	
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers
	<u> </u>

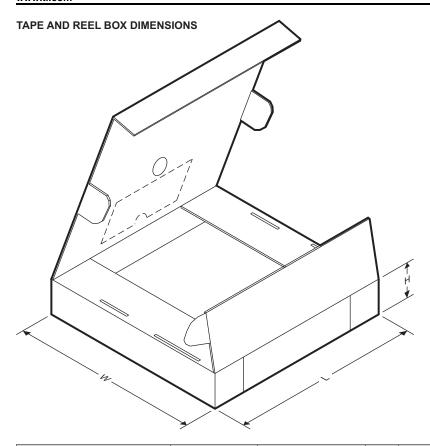
# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMH1208RTVR	WQFN	RTV	32	1000	178.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
LMH1208RTVT	WQFN	RTV	32	250	178.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2

Product Folder Links: LMH1208





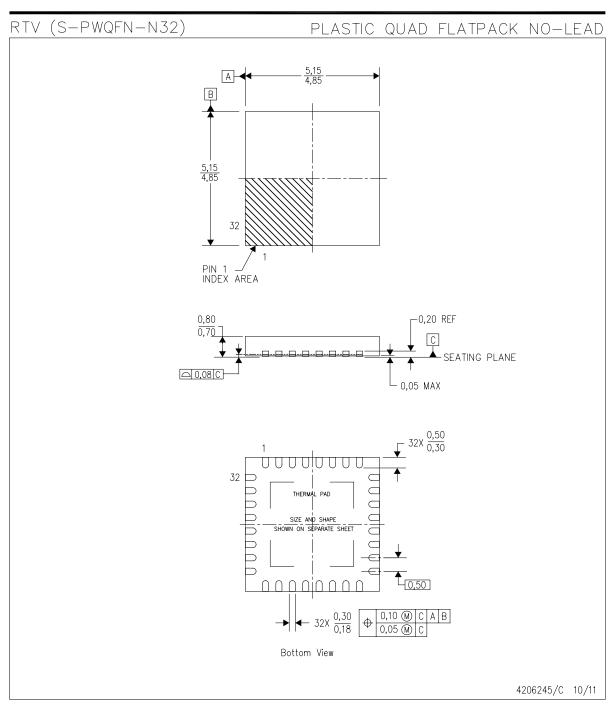
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMH1208RTVR	WQFN	RTV	32	1000	210.0	185.0	35.0
LMH1208RTVT	WQFN	RTV	32	250	210.0	185.0	35.0

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## **MECHANICAL DATA**



NOTES: All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5—1994.

- This drawing is subject to change without notice.
- Quad Flatpack, No-Leads (QFN) package configuration.
- The package thermal pad must be soldered to the board for thermal and mechanical performance.

  See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- Falls within JEDEC MO-220.





#### THERMAL PAD MECHANICAL DATA

# RTV (S-PWQFN-N32)

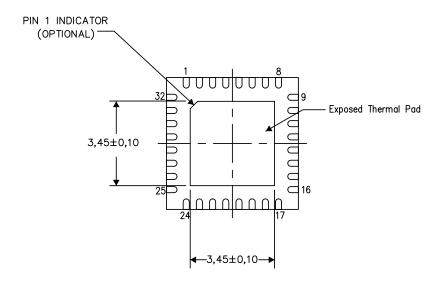
# PLASTIC QUAD FLATPACK NO-LEAD

### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

4206250-2/Q 05/15

NOTE: All linear dimensions are in millimeters

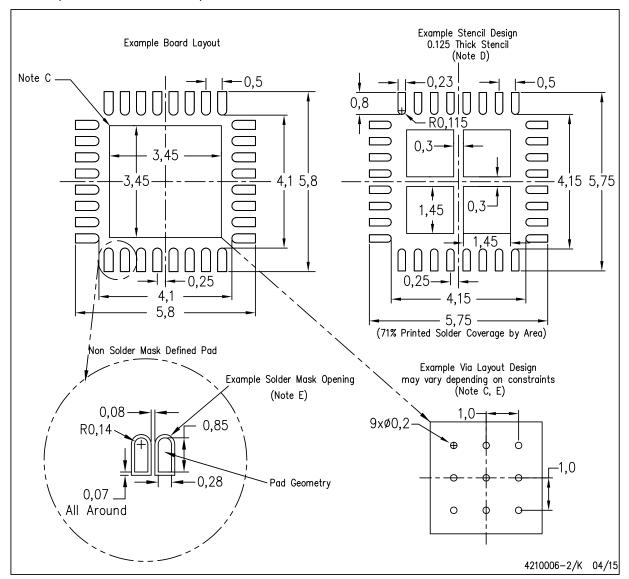




#### LAND PATTERN DATA

# RTV (S-PWQFN-N32)

# PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- 3. This drawing is subject to change without notice.
- C. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">www.ti.com</a>.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- E. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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# PACKAGE OPTION ADDENDUM

10-Dec-2020

#### PACKAGING INFORMATION

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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
LMH1208RTVR	ACTIVE	WQFN	RTV	32	1000	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	L1208	Samples
LMH1208RTVT	ACTIVE	WQFN	RTV	32	250	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	L1208	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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10-Dec-2020

# **PACKAGE MATERIALS INFORMATION**

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### TAPE AND REEL INFORMATION



# TAPE DIMENSIONS + K0 - P1 - B0 W Cavity - A0 -

A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

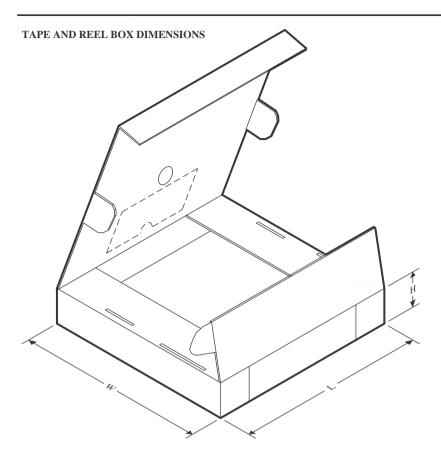


#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMH1208RTVR	WQFN	RTV	32	1000	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
LMH1208RTVT	WQFN	RTV	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2

# **PACKAGE MATERIALS INFORMATION**

www.ti.com 25-Feb-2023

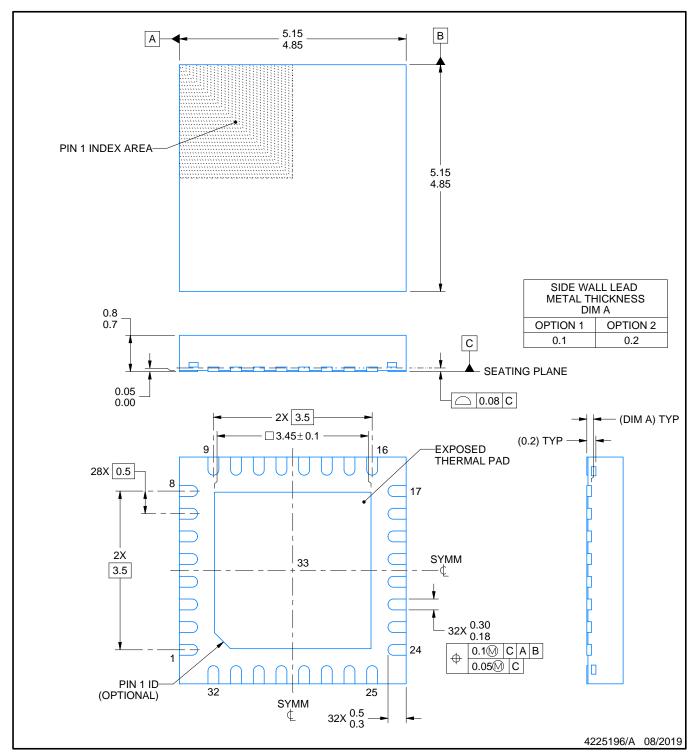


#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMH1208RTVR	WQFN	RTV	32	1000	210.0	185.0	35.0
LMH1208RTVT	WQFN	RTV	32	250	210.0	185.0	35.0



PLASTIC QUAD FLATPACK - NO LEAD

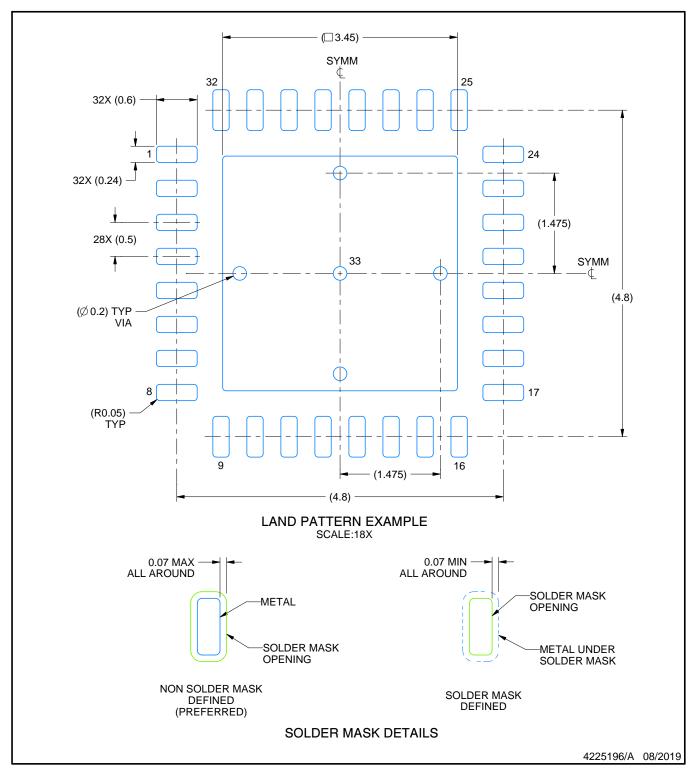


#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
  2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

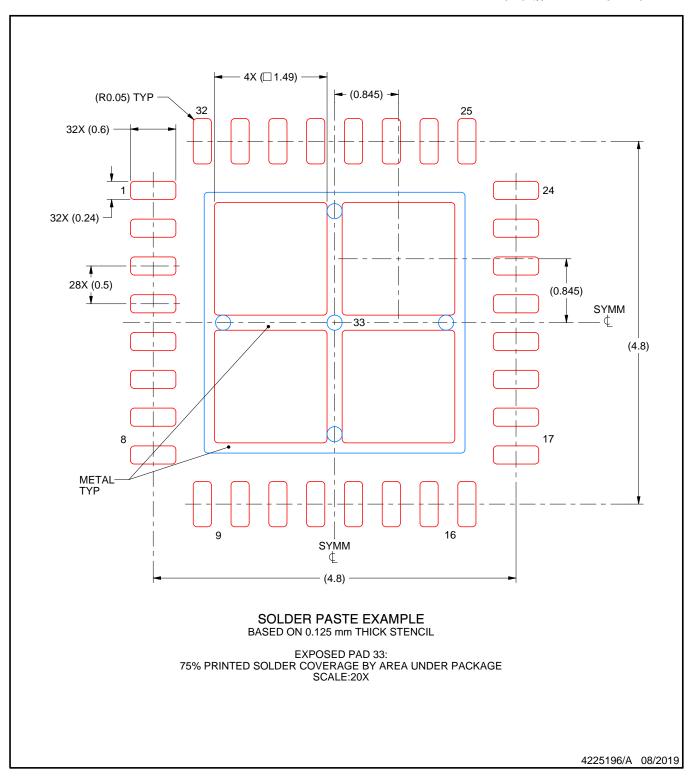


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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