

## ISO7731B High-Speed, Basic Insulation Triple-Channel Digital Isolator

### 1 Features

- 100 Mbps Data Rate
- Robust Isolation Barrier:
  - >100-Year Projected Lifetime at 1 kV<sub>RMS</sub> Working Voltage
  - Up to 5000 V<sub>RMS</sub> Isolation Rating
  - ±100 kV/μs Typical CMTI
- Wide Supply Range: 2.25 V to 5.5 V
- 2.25-V to 5.5-V Level Translation
- Default Output *High* (ISO7731B) and *Low* (ISO7731FB) Options
- Wide Temperature Range: –55°C to +125°C
- Low Power Consumption, Typical 1.7 mA per Channel at 1 Mbps
- Low Propagation Delay: 11 ns Typical (5-V Supplies)
- Robust Electromagnetic Compatibility (EMC)
  - System-Level ESD, EFT, and Surge Immunity
  - ±8 kV IEC 61000-4-2 Contact Discharge Protection across Isolation Barrier
  - Low Emissions
- Wide-SOIC (DW-16) Package
- Safety-Related Certifications:
  - Basic Insulation per DIN V VDE V 0884-11:2017-01
  - UL 1577 Component Recognition Program
  - CSA Certification per IEC 60950-1 and IEC 62368-1 End Equipment Standards
  - CQC Certification per GB4943.1-2011
  - TUV Certification according to EN 60950-1 and EN 61010-1
  - CQC Certification Planned; All Other Certifications Complete

### 2 Applications

- Industrial Automation
- Motor Control
- Power Supplies
- Solar Inverters
- Medical Equipment

### 3 Description

The ISO7731B device is a high-performance, triple-channel digital isolator with 5000 V<sub>RMS</sub> isolation ratings per UL 1577 and 8000 V<sub>PK</sub> basic insulation according to VDE.

The ISO7731B device provides high electromagnetic immunity and low emissions at low power consumption, while isolating CMOS or LVCMOS digital I/Os. Each isolation channel has a logic input and output buffer separated by a double capacitive silicon dioxide (SiO<sub>2</sub>) insulation barrier. This device comes with enable pins which can be used to put the respective outputs in high impedance for multi-master driving applications and to reduce power consumption. The ISO7731B device has two forward and one reverse-direction channels. If the input power or signal is lost, the default output is *high* for device without suffix F and *low* for device with suffix F. See the [Device Functional Modes](#) section for further details.

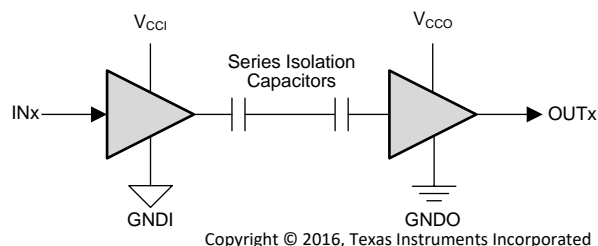
Used in conjunction with isolated power supplies, this device helps prevent noise currents on data buses, such as RS-485, RS-232, and CAN, or other circuits from entering the local ground and interfering with or damaging sensitive circuitry. Through innovative chip design and layout techniques, electromagnetic compatibility of the ISO7731B device has been significantly enhanced to ease system-level ESD, EFT, surge, and emissions compliance. The ISO7731B device is available in 16-pin wide-SOIC package.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
ISO7731B	SOIC (16)	10.30 mm x 7.50 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### Simplified Schematic



V<sub>CCI</sub>=Input V<sub>CC</sub>, V<sub>CCO</sub>=Output V<sub>CC</sub>

GNDI=Input ground, GNDO=Output ground



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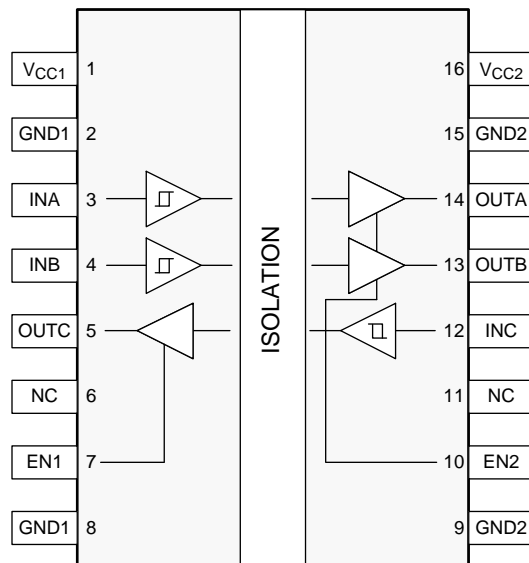
## 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

<b>Changes from Original (April 2018) to Revision A</b>	<b>Page</b>
• Made editorial and cosmetic changes throughout the document .....	1
• Changed From: "Isolation Barrier Life: >40 Years" To: ">100-Year Projected Lifetime at 1 kV <sub>RMS</sub> Working Voltage" in <a href="#">Features</a> .....	1
• Added "Up to 5000 V <sub>RMS</sub> Isolation Rating" in <a href="#">Features</a> .....	1
• Added "±8 kV IEC 61000-4-2 Contact Discharge Protection across Isolation Barrier" in <a href="#">Features</a> .....	1
• Changed From: "All Certifications are Planned" To: "CQC Certification Planned; All Other Certifications Complete" in <a href="#">Features</a> .....	1
• Changed <a href="#">Simplified Schematic</a> to show two isolation capacitors in series instead of a single capacitor .....	1
• Added "Contact discharge per IEC 61000-4-2" specification of ±8000 V in <a href="#">ESD Ratings</a> table .....	4
• Added table note to <i>Data rate</i> specification in <a href="#">Recommended Operating Conditions</a> table .....	4
• Updated certification information in <a href="#">Safety-Related Certifications</a> table .....	7
• Deleted "EN" from <a href="#">Figure 12</a> .....	15
• Updated <a href="#">Figure 15</a> by fixing ground symbols of "Input (Devices with F suffix)" schematic .....	19
• Added " <i>How to use isolation to improve ESD, EFT and Surge immunity in industrial systems</i> " application report in <a href="#">Related Documentation</a> section .....	24

## 5 Pin Configuration and Functions

**ISO7731B DW Package  
16-Pin SOIC  
Top View**



**Pin Functions**

PIN NAME	PIN NO.	I/O	DESCRIPTION
EN1	7	I	Output enable 1. Output pins on side 1 are enabled when EN1 is high or open and in high-impedance state when EN1 is low.
EN2	10	I	Output enable 2. Output pins on side 2 are enabled when EN2 is high or open and in high-impedance state when EN2 is low.
GND1	2, 8	—	Ground connection for $V_{CC1}$
GND2	9, 15	—	Ground connection for $V_{CC2}$
INA	3	I	Input, channel A
INB	4	I	Input, channel B
INC	12	I	Input, channel C
NC	6, 11	—	Not connected
OUTA	14	O	Output, channel A
OUTB	13	O	Output, channel B
OUTC	5	O	Output, channel C
$V_{CC1}$	1	—	Power supply, $V_{CC1}$
$V_{CC2}$	16	—	Power supply, $V_{CC2}$

## 6 Specifications

### 6.1 Absolute Maximum Ratings

 See <sup>(1)</sup>

		MIN	MAX	UNIT
$V_{CC1}, V_{CC2}$	Supply voltage <sup>(2)</sup>	-0.5	6	V
V	Voltage at INx, OUTx, ENx	-0.5	$V_{CCX} + 0.5$ <sup>(3)</sup>	V
$I_O$	Output current	-15	15	mA
$T_J$	Junction temperature		150	°C
$T_{stg}$	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
- (3) Maximum voltage must not exceed 6 V.

### 6.2 ESD Ratings

		VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±6000
		Charged device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1500
		Contact discharge per IEC 61000-4-2; Isolation barrier withstand test <sup>(3)(4)</sup>	±8000

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.
- (3) IEC ESD strike is applied across the barrier with all pins on each side tied together creating a two-terminal device.
- (4) Testing is carried out in air or oil to determine the intrinsic contact discharge capability of the device.

### 6.3 Recommended Operating Conditions

		MIN	NOM	MAX	UNIT
$V_{CC1}, V_{CC2}$	Supply voltage	2.25		5.5	V
$V_{CC(UVLO+)}$	UVLO threshold when supply voltage is rising		2	2.25	V
$V_{CC(UVLO-)}$	UVLO threshold when supply voltage is falling	1.7	1.8		V
$V_{HYS(UVLO)}$	Supply voltage UVLO hysteresis	100	200		mV
$I_{OH}$	High-level output current	$V_{CCO}^{(1)} = 5\text{ V}$		-4	mA
		$V_{CCO} = 3.3\text{ V}$		-2	
		$V_{CCO} = 2.5\text{ V}$		-1	
$I_{OL}$	Low-level output current	$V_{CCO} = 5\text{ V}$		4	mA
		$V_{CCO} = 3.3\text{ V}$		2	
		$V_{CCO} = 2.5\text{ V}$		1	
$V_{IH}$	High-level input voltage	$0.7 \times V_{CCI}^{(1)}$		$V_{CCI}$	V
$V_{IL}$	Low-level input voltage	0		$0.3 \times V_{CCI}$	V
DR <sup>(2)</sup>	Data rate	0		100	Mbps
$T_A$	Ambient temperature	-55	25	125	°C

- (1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ .
- (2) 100 Mbps is the maximum specified data rate, although higher data rates are possible.

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		ISO7731B	
		DW (SOIC)	
		16 PINS	
			UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	81.4	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	44.9	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	45.9	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	28.1	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	45.5	°C/W
$R_{\theta JC(bottom)}$	Junction-to-case(bottom) thermal resistance	—	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics application report](#).

## 6.5 Power Ratings

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$P_D$	Maximum power dissipation	$V_{CC1} = V_{CC2} = 5.5\text{ V}$ , $T_J = 150^\circ\text{C}$ , $C_L = 15\text{ pF}$ , input a 50-MHz 50% duty cycle square wave			150	mW
$P_{D1}$	Maximum power dissipation by side-1	$V_{CC1} = V_{CC2} = 5.5\text{ V}$ , $T_J = 150^\circ\text{C}$ , $C_L = 15\text{ pF}$ , input a 50-MHz 50% duty cycle square wave			50	mW
$P_{D2}$	Maximum power dissipation by side-2	$V_{CC1} = V_{CC2} = 5.5\text{ V}$ , $T_J = 150^\circ\text{C}$ , $C_L = 15\text{ pF}$ , input a 50-MHz 50% duty cycle square wave			100	mW

## 6.6 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
CLR	External clearance <sup>(1)</sup>	Shortest terminal-to-terminal distance through air	>8	mm
CPG	External creepage <sup>(1)</sup>	Shortest terminal-to-terminal distance across the package surface	>8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>21	μm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>600	V
	Material group		I	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 150 V <sub>RMS</sub>	I–IV	
		Rated mains voltage ≤ 300 V <sub>RMS</sub>	I–IV	
		Rated mains voltage ≤ 600 V <sub>RMS</sub>	I–IV	
		Rated mains voltage ≤ 1000 V <sub>RMS</sub>	I–III	
<b>DIN V VDE V 0884-11:2017-01 <sup>(2)</sup></b>				
V <sub>IORM</sub>	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	1414	V <sub>PK</sub>
V <sub>IOWM</sub>	Maximum working isolation voltage	AC voltage (sine wave); time dependent dielectric breakdown (TDDB) test	1000	V <sub>RMS</sub>
		DC voltage	1414	V <sub>DC</sub>
V <sub>IOTM</sub>	Maximum transient isolation voltage	V <sub>TEST</sub> = V <sub>IOTM</sub> = 8000 V <sub>PK</sub> , t = 60 s (qualification); V <sub>TEST</sub> = 1.2 × V <sub>IOTM</sub> = 9600 V <sub>PK</sub> , t = 1 s (100% production)	8000	V <sub>PK</sub>
V <sub>IOSM</sub>	Maximum surge isolation voltage <sup>(3)</sup>	Test method per IEC 62368-1, 1.2/50 μs waveform, V <sub>TEST</sub> = 1.3 × V <sub>IOSM</sub> = 7800 V <sub>PK</sub> (qualification)	6000	V <sub>PK</sub>
q <sub>pd</sub>	Apparent charge <sup>(4)</sup>	Method a, After I/O safety test subgroup 2/3, V <sub>ini</sub> = V <sub>IOTM</sub> , t <sub>ini</sub> = 60 s; V <sub>pd(m)</sub> = 1.2 × V <sub>IORM</sub> = 1697 V <sub>PK</sub> , t <sub>m</sub> = 10 s	≤5	pC
		Method a, After environmental tests subgroup 1, V <sub>ini</sub> = V <sub>IOTM</sub> , t <sub>ini</sub> = 60 s; V <sub>pd(m)</sub> = 1.2 × V <sub>IORM</sub> = 1697 V <sub>PK</sub> , t <sub>m</sub> = 10 s	≤5	
		Method b1; At routine test (100% production) and preconditioning (type test) V <sub>ini</sub> = 1.2 × V <sub>IOTM</sub> = 9600 V <sub>PK</sub> , t <sub>ini</sub> = 1 s; V <sub>pd(m)</sub> = 1.5 × V <sub>IORM</sub> = 2121 V <sub>PK</sub> , t <sub>m</sub> = 1 s	≤5	
C <sub>IO</sub>	Barrier capacitance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 0.4 sin (2πft), f = 1 MHz	~0.7	pF
R <sub>IO</sub>	Isolation resistance <sup>(5)</sup>	V <sub>IO</sub> = 500 V, T <sub>A</sub> = 25°C	>10 <sup>12</sup>	Ω
		V <sub>IO</sub> = 500 V, 100°C ≤ T <sub>A</sub> ≤ 125°C	>10 <sup>11</sup>	
		V <sub>IO</sub> = 500 V at T <sub>S</sub> = 150°C	>10 <sup>9</sup>	
	Pollution degree		2	
	Climatic category		55/125/21	
<b>UL 1577</b>				
V <sub>ISO</sub>	Withstanding isolation voltage	V <sub>TEST</sub> = V <sub>ISO</sub> = 5000 V <sub>RMS</sub> , t = 60 s (qualification), V <sub>TEST</sub> = 1.2 × V <sub>ISO</sub> = 6000 V <sub>RMS</sub> , t = 1 s (100% production)	5000	V <sub>RMS</sub>

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *basic electrical insulation* only within the maximum operating ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.
- (4) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (5) All pins on each side of the barrier tied together creating a two-terminal device.

## 6.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Certified according to DIN V VDE V 0884-11:2017-01 and DIN EN 61010-1 (VDE 0411-1):2011-07	Certified according to IEC 60950-1 and IEC 62368-1	Recognized under UL 1577 Component Recognition Program	Plan to certify according to GB 4943.1-2011	Certified according to EN 61010-1:2010 (3rd Ed) and EN 60950-1:2006/A2:2013
Basic Insulation; Maximum transient isolation voltage, 8000 V <sub>PK</sub> ; Maximum repetitive peak isolation voltage, 1414 V <sub>PK</sub> ; Maximum surge isolation voltage, 6000 V <sub>PK</sub>	Basic Insulation; 800 V <sub>RMS</sub> working voltage per CSA 60950-1-07+A1+A2, IEC 60950-1 2nd Ed. +A1+A2, CSA 62368-1-14 and IEC 62368-1:2014	Single protection, 5000 V <sub>RMS</sub>	700 V <sub>RMS</sub> maximum working voltage, Altitude ≤ 5000 m, Tropical Climate	5000 V <sub>RMS</sub> insulation per EN 61010-1:2010 (3rd Ed) up to working voltage of 600 V <sub>RMS</sub> 5000 V <sub>RMS</sub> insulation per EN 60950-1:2006/A2:2013 up to working voltage of 800 V <sub>RMS</sub>
Certificate number: 40047657	Master contract number: 220991	File number: E181974	Certification planned	Client ID number: 77311

## 6.8 Safety Limiting Values

Safety limiting intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>S</sub>	Safety input, output, or supply current <sup>(1)</sup>	R <sub>θJA</sub> = 81.4 °C/W, V <sub>I</sub> = 5.5 V, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C, see <a href="#">Figure 1</a>			279	mA
		R <sub>θJA</sub> = 81.4 °C/W, V <sub>I</sub> = 3.6 V, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C, see <a href="#">Figure 1</a>			427	
		R <sub>θJA</sub> = 81.4 °C/W, V <sub>I</sub> = 2.75 V, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C, see <a href="#">Figure 1</a>			558	
P <sub>S</sub>	Safety input, output, or total power <sup>(1)</sup>	R <sub>θJA</sub> = 81.4 °C/W, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C, see <a href="#">Figure 2</a>			1536	mW
T <sub>S</sub>	Maximum safety temperature <sup>(1)</sup>				150	°C

- (1) The maximum safety temperature, T<sub>S</sub>, has the same value as the maximum junction temperature, T<sub>J</sub>, specified for the device. The I<sub>S</sub> and P<sub>S</sub> parameters represent the safety current and safety power respectively. The maximum limits of I<sub>S</sub> and P<sub>S</sub> should not be exceeded. These limits vary with the ambient temperature, T<sub>A</sub>.

The junction-to-air thermal resistance, R<sub>θJA</sub>, in the [Thermal Information](#) table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

$$T_J = T_A + R_{\theta JA} \times P, \text{ where } P \text{ is the power dissipated in the device.}$$

$$T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S, \text{ where } T_{J(max)} \text{ is the maximum allowed junction temperature.}$$

$$P_S = I_S \times V_I, \text{ where } V_I \text{ is the maximum input voltage.}$$

## 6.9 Electrical Characteristics—5-V Supply

 $V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OH}$	High-level output voltage $I_{OH} = -4\text{ mA}$ ; see <a href="#">Figure 9</a>	$V_{CCO}^{(1)} - 0.4$	4.8		V
$V_{OL}$	Low-level output voltage $I_{OL} = 4\text{ mA}$ ; see <a href="#">Figure 9</a>		0.2	0.4	V
$V_{IT+(IN)}$	Rising input voltage threshold		$0.6 \times V_{CCI}$	$0.7 \times V_{CCI}$	V
$V_{IT-(IN)}$	Falling input voltage threshold	$0.3 \times V_{CCI}$	$0.4 \times V_{CCI}$		V
$V_{I(HYS)}$	Input threshold voltage hysteresis	$0.1 \times V_{CCI}$	$0.2 \times V_{CCI}$		V
$I_{IH}$	High-level input current $V_{IH} = V_{CCI}^{(1)}$ at INx or ENx			10	$\mu\text{A}$
$I_{IL}$	Low-level input current $V_{IL} = 0\text{ V}$ at INx or ENx	-10			$\mu\text{A}$
$ CM_H $	High-level common-mode transient immunity $V_I = V_{CCI}^{(2)}$ , $V_{CM} = 1200\text{ V}$ ; see <a href="#">Figure 12</a>	85	100		$\text{kV}/\mu\text{s}$
$ CM_L $	Low-level common-mode transient immunity $V_I = 0\text{ V}$ , $V_{CM} = 1200\text{ V}$ ; see <a href="#">Figure 12</a>	85	100		$\text{kV}/\mu\text{s}$
$C_I$	Input Capacitance $V_I = V_{CC}^{(2)} / 2 + 0.4 \times \sin(2\pi ft)$ , $f = 1\text{ MHz}$ , $V_{CC} = 5\text{ V}$		2		pF

- (1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ .  
 (2) Measured from input pin to ground.

## 6.10 Supply Current Characteristics—5-V Supply

 $V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	SUPPLY CURRENT	MIN	TYP	MAX	UNIT	
Supply current - disable	$EN1 = EN2 = 0\text{ V}$ ; $V_I = V_{CCI}^{(1)}$ (ISO7731B); $V_I = 0\text{ V}$ (ISO7731B with F suffix)	$I_{CC1}$		0.8	1.2	mA	
		$I_{CC2}$		0.7	1	mA	
	$EN1 = EN2 = 0\text{ V}$ ; $V_I = 0\text{ V}$ (ISO7731B); $V_I = V_{CCI}$ (ISO7731B with F suffix)	$I_{CC1}$		3	4.3	mA	
		$I_{CC2}$		1.8	2.6	mA	
Supply current - DC signal	$EN1 = EN2 = V_{CCI}$ ; $V_I = V_{CCI}$ (ISO7731B); $V_I = 0\text{ V}$ (ISO7731B with F suffix)	$I_{CC1}$		1.3	1.7	mA	
		$I_{CC2}$		1.6	2.2	mA	
	$EN1 = EN2 = V_{CCI}$ ; $V_I = 0\text{ V}$ (ISO7731B); $V_I = V_{CCI}$ (ISO7731B with F suffix)	$I_{CC1}$		3.5	5	mA	
		$I_{CC2}$		2.8	4.1	mA	
Supply current - AC signal	$EN1 = EN2 = V_{CCI}$ ; All channels switching with square wave clock input; $C_L = 15\text{ pF}$	1 Mbps	$I_{CC1}$		2.7	3.4	mA
			$I_{CC2}$		2.3	3.3	mA
		10 Mbps	$I_{CC1}$		3	4	mA
			$I_{CC2}$		3.3	4.4	mA
		100 Mbps	$I_{CC1}$		8.5	11	mA
			$I_{CC2}$		13.1	16	mA

- (1)  $V_{CCI}$  = Input-side  $V_{CC}$



## 6.11 Electrical Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OH}$	High-level output voltage $I_{OH} = -2\text{ mA}$ ; see <a href="#">Figure 9</a>	$V_{CCO}^{(1)} - 0.3$	3.2		V
$V_{OL}$	Low-level output voltage $I_{OL} = 2\text{ mA}$ ; see <a href="#">Figure 9</a>		0.1	0.3	V
$V_{IT+(IN)}$	Rising input voltage threshold		$0.6 \times V_{CCI}$	$0.7 \times V_{CCI}$	V
$V_{IT-(IN)}$	Falling input voltage threshold	$0.3 \times V_{CCI}$	$0.4 \times V_{CCI}$		V
$V_{I(HYS)}$	Input threshold voltage hysteresis	$0.1 \times V_{CCI}$	$0.2 \times V_{CCI}$		V
$I_{IH}$	High-level input current $V_{IH} = V_{CCI}^{(1)}$ at INx or ENx			10	$\mu\text{A}$
$I_{IL}$	Low-level input current $V_{IL} = 0\text{ V}$ at INx or ENx	-10			$\mu\text{A}$
$ CM_H $	High-level common-mode transient immunity $V_I = V_{CCI}^{(1)}$ , $V_{CM} = 1200\text{ V}$ ; see <a href="#">Figure 12</a>	85	100		$\text{kV}/\mu\text{s}$
$ CM_L $	Low-level common-mode transient immunity $V_I = 0\text{ V}$ , $V_{CM} = 1200\text{ V}$ ; see <a href="#">Figure 12</a>	85	100		$\text{kV}/\mu\text{s}$

(1)  $V_{CC1}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ .

## 6.12 Supply Current Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	SUPPLY CURRENT	MIN	TYP	MAX	UNIT	
Supply current - disable	$EN1 = EN2 = 0\text{ V}$ ; $V_I = V_{CCI}^{(1)}$ (ISO7731B); $V_I = 0\text{ V}$ (ISO7731B with F suffix)	$I_{CC1}$		0.8	1.2	$\text{mA}$	
		$I_{CC2}$		0.7	1	$\text{mA}$	
	$EN1 = EN2 = 0\text{ V}$ ; $V_I = 0\text{ V}$ (ISO7731B); $V_I = V_{CCI}$ (ISO7731B with F suffix)	$I_{CC1}$		3	4.3	$\text{mA}$	
		$I_{CC2}$		1.8	2.6	$\text{mA}$	
Supply current - DC signal	$EN1 = EN2 = V_{CCI}$ ; $V_I = V_{CCI}$ (ISO7731B); $V_I = 0\text{ V}$ (ISO7731B with F suffix)	$I_{CC1}$		1.3	1.7	$\text{mA}$	
		$I_{CC2}$		1.6	2.2	$\text{mA}$	
	$EN1 = EN2 = V_{CCI}$ ; $V_I = 0\text{ V}$ (ISO7731B); $V_I = V_{CCI}$ (ISO7731B with F suffix)	$I_{CC1}$		3.5	5	$\text{mA}$	
		$I_{CC2}$		2.8	4.1	$\text{mA}$	
Supply current - AC signal	$EN1 = EN2 = V_{CCI}$ ; All channels switching with square wave clock input; $C_L = 15\text{ pF}$	1 Mbps	$I_{CC1}$		2.4	3.4	$\text{mA}$
			$I_{CC2}$		2.2	3.3	$\text{mA}$
		10 Mbps	$I_{CC1}$		2.8	3.8	$\text{mA}$
			$I_{CC2}$		2.9	4	$\text{mA}$
		100 Mbps	$I_{CC1}$		6.7	8.5	$\text{mA}$
			$I_{CC2}$		10	12.5	$\text{mA}$

(1)  $V_{CCI}$  = Input-side  $V_{CC}$

### 6.13 Electrical Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OH}$	High-level output voltage	$I_{OH} = -1\text{ mA}$ ; see <a href="#">Figure 9</a>	$V_{CCO}^{(1)} - 0.2$	2.45		V
$V_{OL}$	Low-level output voltage	$I_{OL} = 1\text{ mA}$ ; see <a href="#">Figure 9</a>		0.05	0.2	V
$V_{IT+(IN)}$	Rising input voltage threshold			$0.6 \times V_{CCI}$	$0.7 \times V_{CCI}$	V
$V_{IT-(IN)}$	Falling input voltage threshold		$0.3 \times V_{CCI}$	$0.4 \times V_{CCI}$		V
$V_{I(HYS)}$	Input threshold voltage hysteresis		$0.1 \times V_{CCI}$	$0.2 \times V_{CCI}$		V
$I_{IH}$	High-level input current	$V_{IH} = V_{CCI}^{(1)}$ at INx or ENx			10	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_{IL} = 0\text{ V}$ at INx or ENx	-10			$\mu\text{A}$
$ CM_H $	High-level common-mode transient immunity	$V_I = V_{CCI}^{(1)}$ , $V_{CM} = 1200\text{ V}$ ; see <a href="#">Figure 12</a>	85	100		$\text{kV}/\mu\text{s}$
$ CM_L $	Low-level common-mode transient immunity	$V_I = 0\text{ V}$ , $V_{CM} = 1200\text{ V}$ ; see <a href="#">Figure 12</a>	85	100		$\text{kV}/\mu\text{s}$

 (1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ .

### 6.14 Supply Current Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	SUPPLY CURRENT	MIN	TYP	MAX	UNIT	
Supply current - disable	$EN1 = EN2 = 0\text{ V}$ ; $V_I = V_{CCI}^{(1)}$ (ISO7731B); $V_I = 0\text{ V}$ (ISO7731B with F suffix)	$I_{CC1}$		0.8	1.2	$\text{mA}$	
		$I_{CC2}$		0.7	1	$\text{mA}$	
	$EN1 = EN2 = 0\text{ V}$ ; $V_I = 0\text{ V}$ (ISO7731B); $V_I = V_{CCI}$ (ISO7731B with F suffix)	$I_{CC1}$		3	4.3	$\text{mA}$	
		$I_{CC2}$		1.8	2.6	$\text{mA}$	
Supply current - DC signal	$EN1 = EN2 = V_{CCI}$ ; $V_I = V_{CCI}$ (ISO7731B); $V_I = 0\text{ V}$ (ISO7731B with F suffix)	$I_{CC1}$		1.3	1.7	$\text{mA}$	
		$I_{CC2}$		1.6	2.2	$\text{mA}$	
	$EN1 = EN2 = V_{CCI}$ ; $V_I = 0\text{ V}$ (ISO7731B); $V_I = V_{CCI}$ (ISO7731B with F suffix)	$I_{CC1}$		3.5	5	$\text{mA}$	
		$I_{CC2}$		2.8	4.1	$\text{mA}$	
Supply current - AC signal	$EN1 = EN2 = V_{CCI}$ ; All channels switching with square wave clock input; $C_L = 15\text{ pF}$	1 Mbps	$I_{CC1}$		2.4	3.4	$\text{mA}$
			$I_{CC2}$		2.2	3.2	$\text{mA}$
		10 Mbps	$I_{CC1}$		2.7	3.7	$\text{mA}$
			$I_{CC2}$		2.7	3.8	$\text{mA}$
		100 Mbps	$I_{CC1}$		5.6	7	$\text{mA}$
			$I_{CC2}$		8	10	$\text{mA}$

 (1)  $V_{CCI}$  = Input-side  $V_{CC}$

## 6.15 Switching Characteristics—5-V Supply

 $V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ , $t_{PHL}$	Propagation delay time	See Figure 9	6	11	16	ns
PWD	Pulse width distortion <sup>(1)</sup> $ t_{PHL} - t_{PLH} $		0.6	4.9	ns	
$t_{sk(o)}$	Channel-to-channel output skew time <sup>(2)</sup>	Same-direction channels			4	ns
$t_{sk(pp)}$	Part-to-part skew time <sup>(3)</sup>				4.5	ns
$t_r$	Output signal rise time	See Figure 9		1.3	3.9	ns
$t_f$	Output signal fall time			1.4	3.9	ns
$t_{PHZ}$	Disable propagation delay, high-to-high impedance output	See Figure 10		8	20	ns
$t_{PLZ}$	Disable propagation delay, low-to-high impedance output			8	20	ns
$t_{PZH}$	Enable propagation delay, high impedance-to-high output for ISO7731B			7	20	ns
	Enable propagation delay, high impedance-to-high output for ISO7731B with F suffix			3	8.5	$\mu\text{s}$
$t_{PZL}$	Enable propagation delay, high impedance-to-low output for ISO7731B			3	8.5	$\mu\text{s}$
	Enable propagation delay, high impedance-to-low output for ISO7731B with F suffix			7	20	ns
$t_{DO}$	Default output delay time from input power loss	Measured from the time $V_{CC}$ goes below 1.7 V. See Figure 11		0.1	0.3	$\mu\text{s}$
$t_{ie}$	Time interval error	$2^{16} - 1$ PRBS data at 100 Mbps		0.6		ns

- (1) Also known as pulse skew.
- (2)  $t_{sk(o)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
- (3)  $t_{sk(pp)}$  is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

## 6.16 Switching Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ , $t_{PHL}$	Propagation delay time	See Figure 9	6	11	16	ns
PWD	Pulse width distortion <sup>(1)</sup> $ t_{PHL} - t_{PLH} $		0.1	5	ns	
$t_{sk(o)}$	Channel-to-channel output skew time <sup>(2)</sup>	Same-direction channels			4.1	ns
$t_{sk(pp)}$	Part-to-part skew time <sup>(3)</sup>				4.5	ns
$t_r$	Output signal rise time	See Figure 9		1.3	3	ns
$t_f$	Output signal fall time			1.3	3	ns
$t_{PHZ}$	Disable propagation delay, high-to-high impedance output	See Figure 10		17	30	ns
$t_{PLZ}$	Disable propagation delay, low-to-high impedance output			17	30	ns
$t_{PZH}$	Enable propagation delay, high impedance-to-high output for ISO7731B			17	30	ns
	Enable propagation delay, high impedance-to-high output for ISO7731B with F suffix			3.2	8.5	$\mu\text{s}$
$t_{PZL}$	Enable propagation delay, high impedance-to-low output for ISO7731B			3.2	8.5	$\mu\text{s}$
	Enable propagation delay, high impedance-to-low output for ISO7731B with F suffix			17	30	ns
$t_{DO}$	Default output delay time from input power loss	Measured from the time $V_{CC}$ goes below 1.7 V. See Figure 11		0.1	0.3	$\mu\text{s}$
$t_{ie}$	Time interval error	$2^{16} - 1$ PRBS data at 100 Mbps		0.6		ns

- (1) Also known as Pulse Skew.
- (2)  $t_{sk(o)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
- (3)  $t_{sk(pp)}$  is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

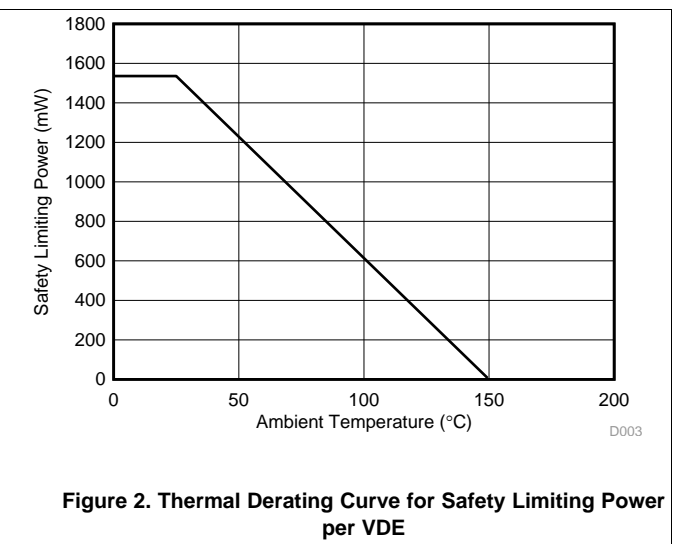
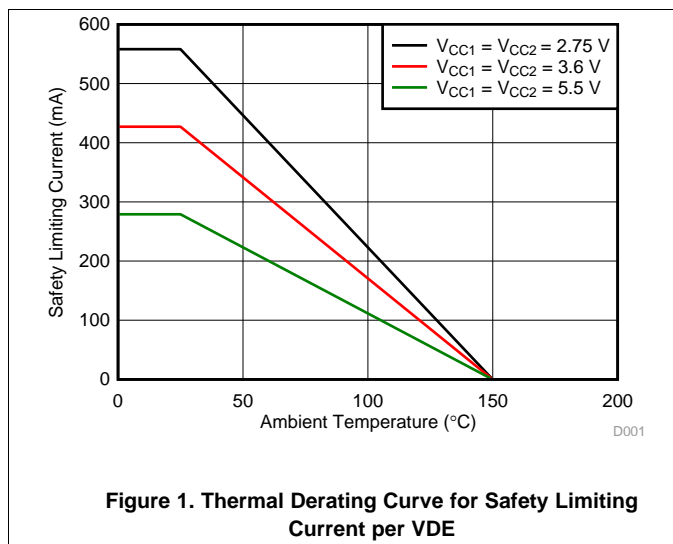
### 6.17 Switching Characteristics—2.5-V Supply

$V_{CC1} = V_{CC2} = 2.5\text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}, t_{PHL}$	Propagation delay time	See Figure 9	7.5	12	18.5	ns
PWD	Pulse width distortion <sup>(1)</sup> $ t_{PHL} - t_{PLH} $		0.2	5.1	ns	
$t_{sk(o)}$	Channel-to-channel output skew time <sup>(2)</sup>	Same-direction Channels			4.1	ns
$t_{sk(pp)}$	Part-to-part skew time <sup>(3)</sup>				4.6	ns
$t_r$	Output signal rise time	See Figure 9		1	3.5	ns
$t_f$	Output signal fall time			1	3.5	ns
$t_{PHZ}$	Disable propagation delay, high-to-high impedance output	See Figure 10		22	40	ns
$t_{PLZ}$	Disable propagation delay, low-to-high impedance output			22	40	ns
$t_{PZH}$	Enable propagation delay, high impedance-to-high output for ISO7731B			18	40	ns
	Enable propagation delay, high impedance-to-high output for ISO7731B with F suffix			3.3	8.5	$\mu\text{s}$
$t_{PZL}$	Enable propagation delay, high impedance-to-low output for ISO7731B			3.3	8.5	$\mu\text{s}$
	Enable propagation delay, high impedance-to-low output for ISO7731B with F suffix			18	40	ns
$t_{DO}$	Default output delay time from input power loss	Measured from the time $V_{CC}$ goes below 1.7 V. See Figure 11		0.1	0.3	$\mu\text{s}$
$t_{ie}$	Time interval error	$2^{16} - 1$ PRBS data at 100 Mbps		0.6		ns

- (1) Also known as pulse skew.
- (2)  $t_{sk(o)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
- (3)  $t_{sk(pp)}$  is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

### 6.18 Insulation Characteristics Curves



6.19 Typical Characteristics

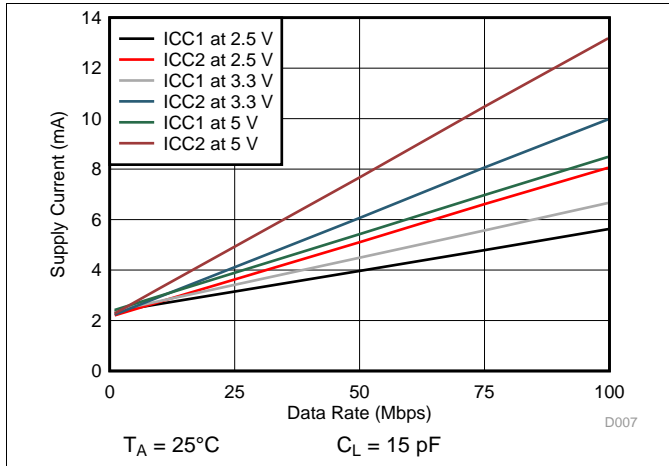


Figure 3. ISO7731B Supply Current vs Data Rate (With 15-pF Load)

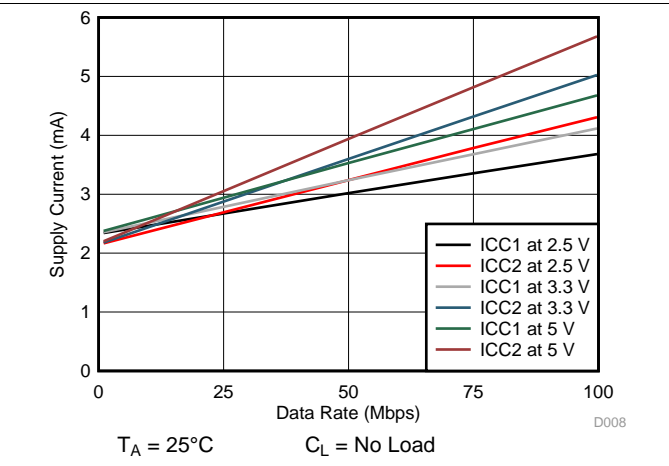


Figure 4. ISO7731B Supply Current vs Data Rate (With No Load)

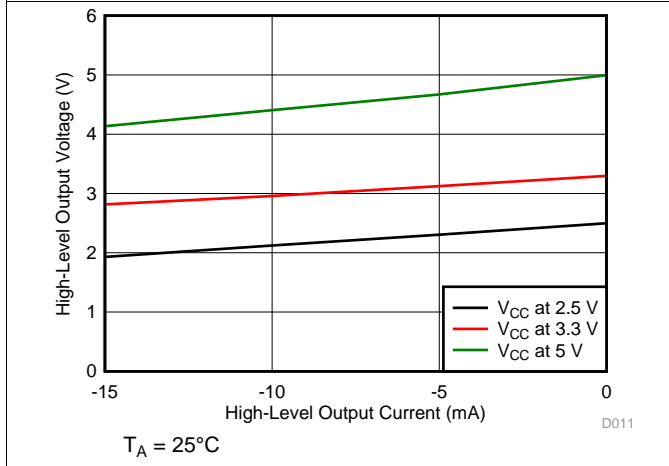


Figure 5. High-Level Output Voltage vs High-level Output Current

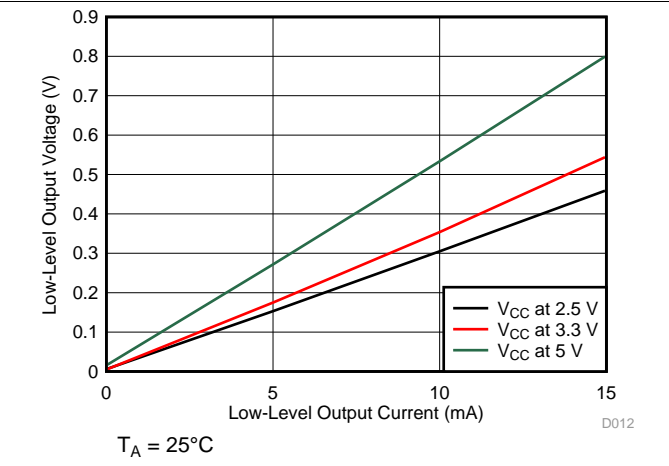


Figure 6. Low-Level Output Voltage vs Low-Level Output Current

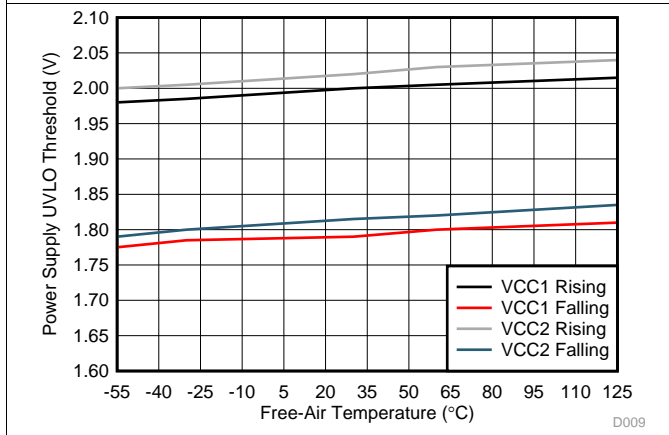


Figure 7. Power Supply Undervoltage Threshold vs Free-Air Temperature

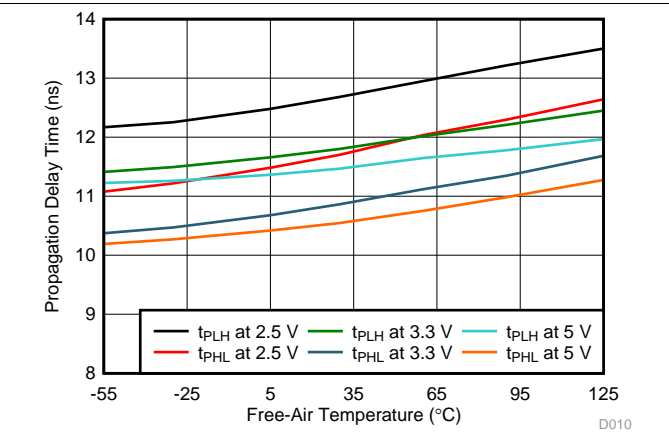
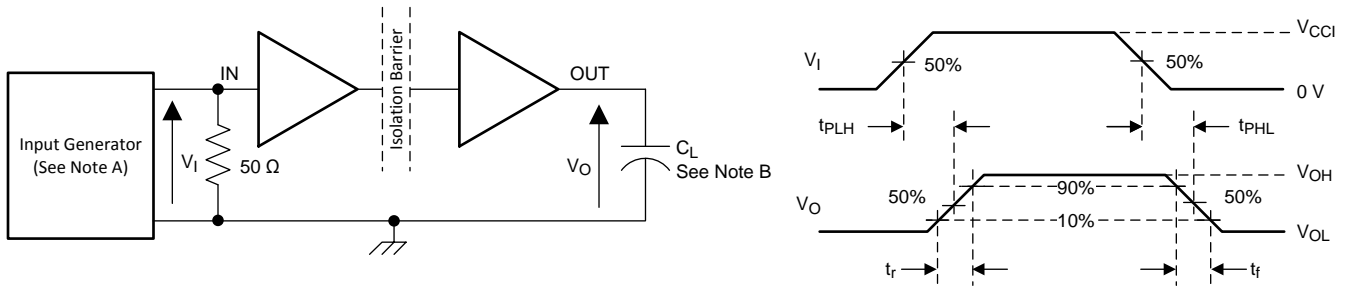


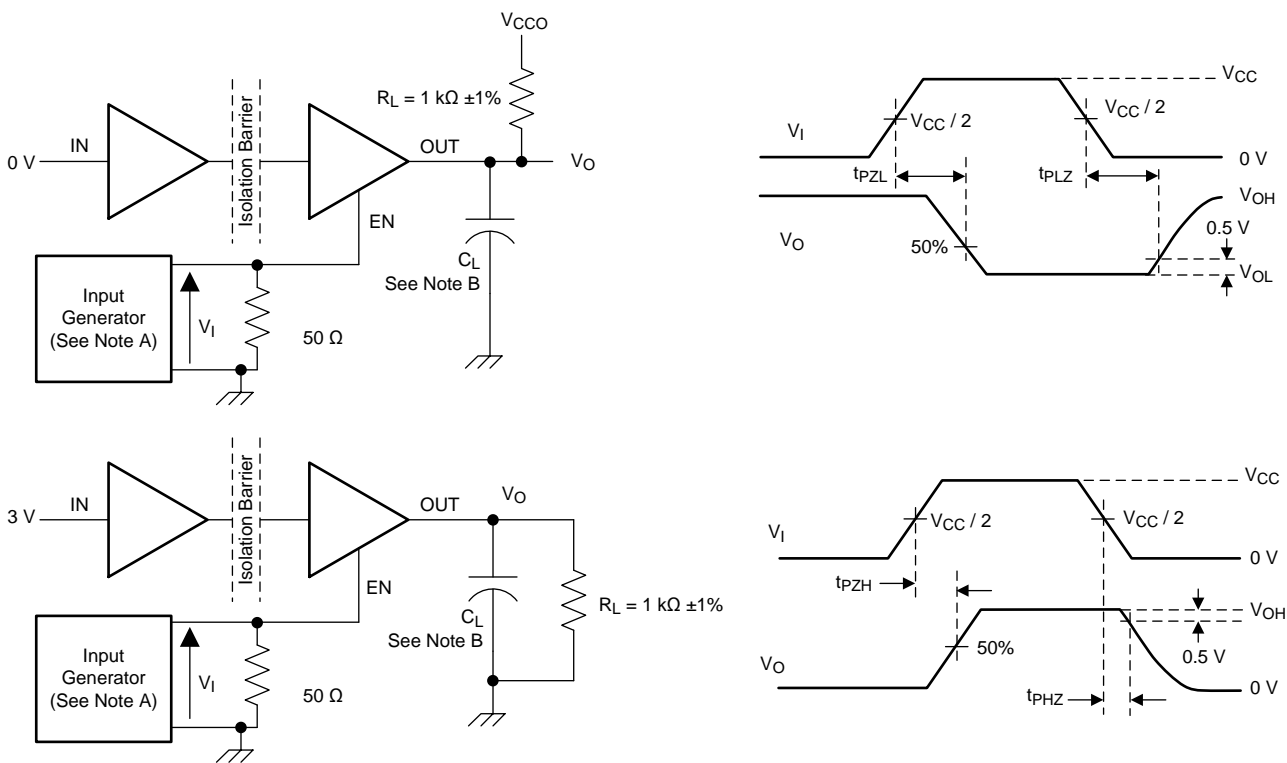
Figure 8. Propagation Delay Time vs Free-Air Temperature

## 7 Parameter Measurement Information



- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  50 kHz, 50% duty cycle,  $t_r \leq 3$  ns,  $t_f \leq 3$  ns,  $Z_O = 50 \Omega$ . At the input,  $50 \Omega$  resistor is required to terminate Input Generator signal. It is not needed in actual application.
- B.  $C_L = 15$  pF and includes instrumentation and fixture capacitance within  $\pm 20\%$ .

**Figure 9. Switching Characteristics Test Circuit and Voltage Waveforms**

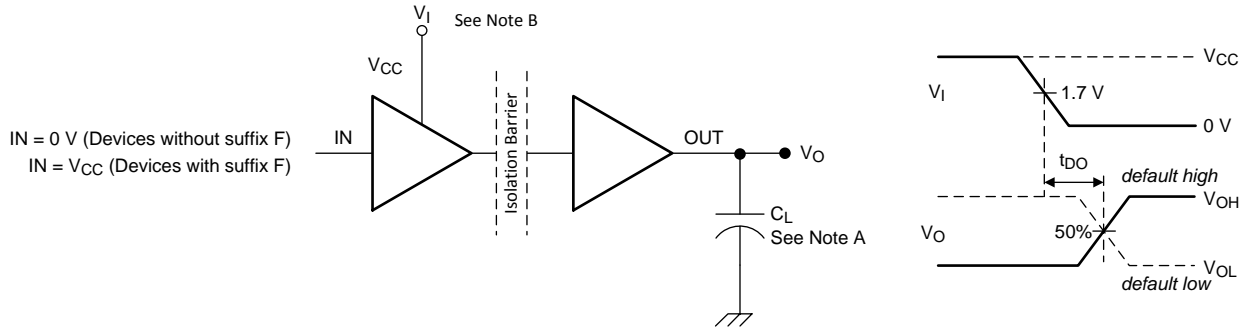


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- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  10 kHz, 50% duty cycle,  $t_r \leq 3$  ns,  $t_f \leq 3$  ns,  $Z_O = 50 \Omega$ .
- B.  $C_L = 15$  pF and includes instrumentation and fixture capacitance within  $\pm 20\%$ .

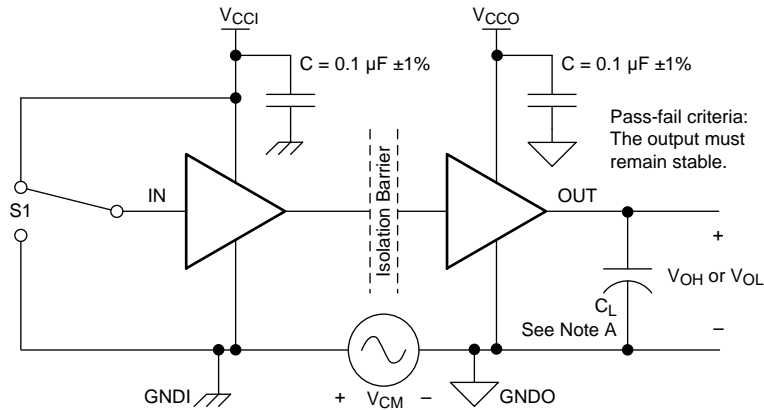
**Figure 10. Enable/Disable Propagation Delay Time Test Circuit and Waveform**

**Parameter Measurement Information (continued)**



- A. C<sub>L</sub> = 15 pF and includes instrumentation and fixture capacitance within ±20%.
- B. Power Supply Ramp Rate = 10 mV/ns

**Figure 11. Default Output Delay Time Test Circuit and Voltage Waveforms**



- A. C<sub>L</sub> = 15 pF and includes instrumentation and fixture capacitance within ±20%.

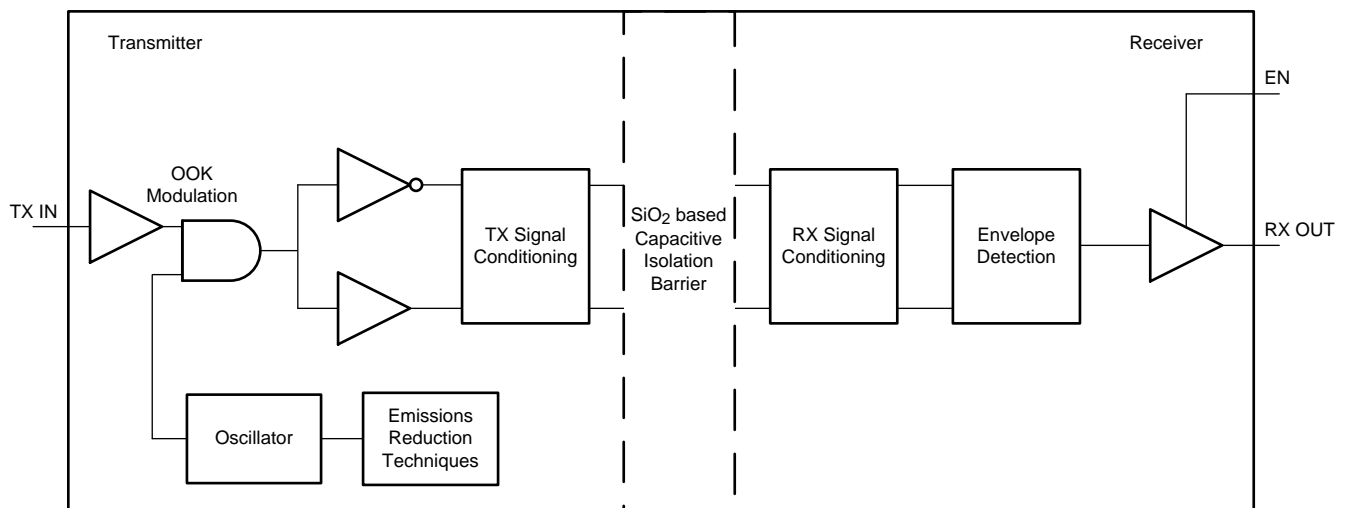
**Figure 12. Common-Mode Transient Immunity Test Circuit**

## 8 Detailed Description

### 8.1 Overview

The ISO7731B device has an ON-OFF keying (OOK) modulation scheme to transmit the digital data across a silicon dioxide based isolation barrier. The transmitter sends a high frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output through a buffer stage. If the ENx pin is low then the output goes to high impedance. The ISO7731B device also incorporates advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions because of the high frequency carrier and IO buffer switching. The conceptual block diagram of a digital capacitive isolator, [Figure 13](#), shows a functional block diagram of a typical channel.

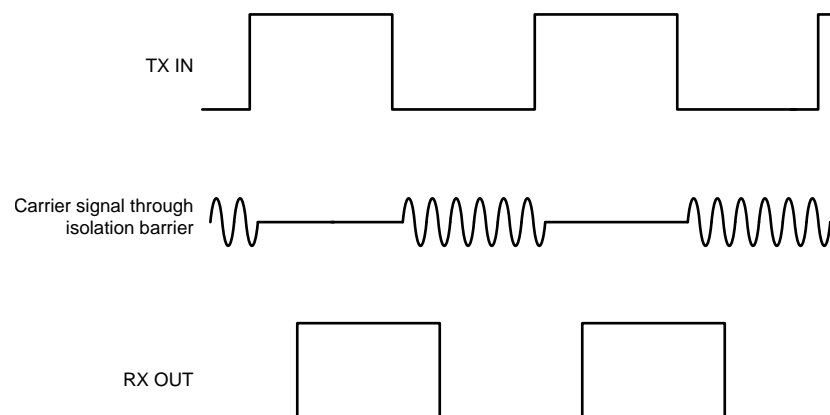
### 8.2 Functional Block Diagram



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**Figure 13. Conceptual Block Diagram of a Digital Capacitive Isolator**

[Figure 14](#) shows a conceptual detail of how the ON-OFF keying scheme works.



**Figure 14. On-Off Keying (OOK) Based Modulation Scheme**



### 8.3 Feature Description

Table 1 provides an overview of the device features.

**Table 1. Device Features**

PART NUMBER	CHANNEL DIRECTION	MAXIMUM DATA RATE	DEFAULT OUTPUT	ISOLATION RATING <sup>(1)</sup>
ISO7731B	2 Forward, 1 Reverse	100 Mbps	High	5000 V <sub>RMS</sub> / 8000 V <sub>PK</sub>
ISO7731B with F suffix	2 Forward, 1 Reverse	100 Mbps	Low	5000 V <sub>RMS</sub> / 8000 V <sub>PK</sub>

(1) See [Safety-Related Certifications](#) for detailed isolation ratings.

#### 8.3.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 22. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO7731B device incorporates many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection cells for input and output signal pins and inter-chip bond pads.
- Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- Reduced common mode currents across the isolation barrier by ensuring purely differential internal operation.

## 8.4 Device Functional Modes

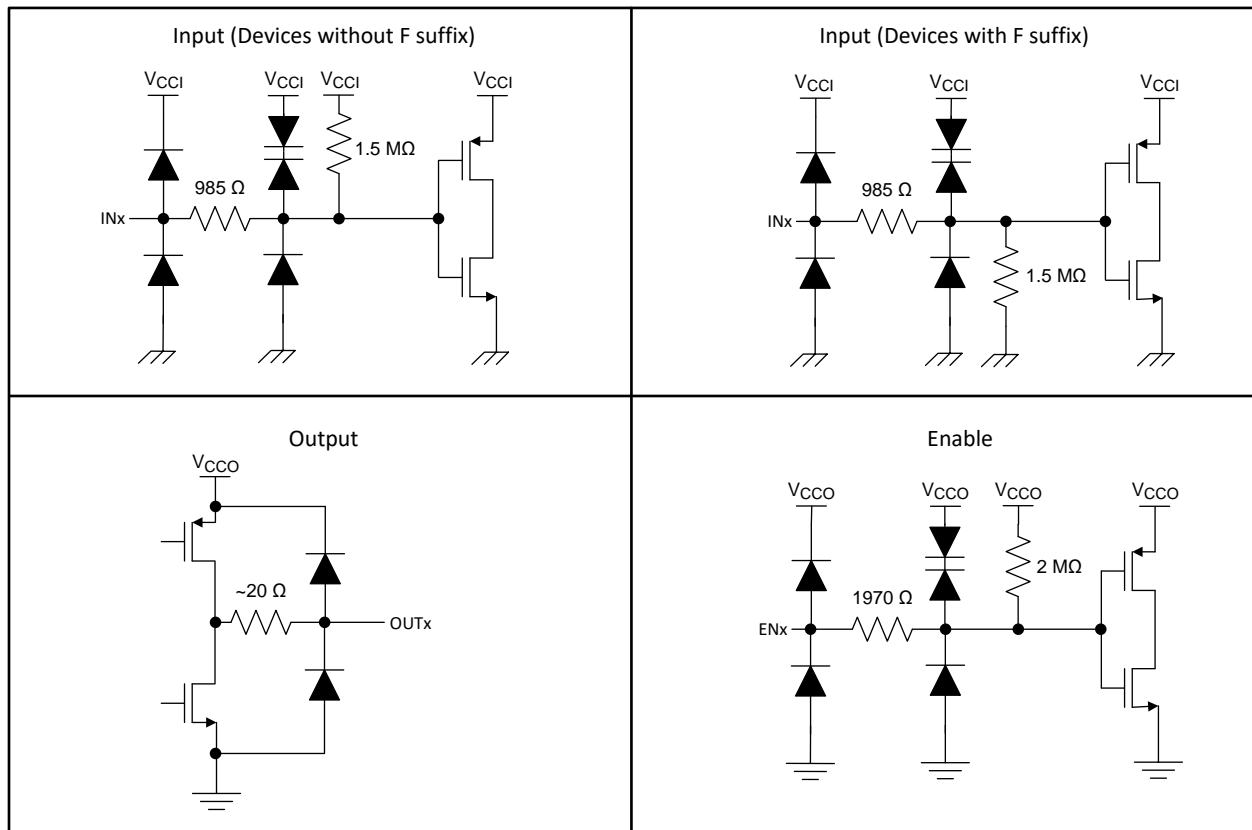
Table 2 lists the functional modes for the ISO7731B device.

**Table 2. Function Table<sup>(1)</sup>**

$V_{CCI}$	$V_{CCO}$	INPUT (INx) <sup>(2)</sup>	OUTPUT ENABLE (ENx)	OUTPUT (OUTx)	COMMENTS
PU	PU	H	H or open	H	Normal Operation: A channel output assumes the logic state of its input.
		L	H or open	L	
		Open	H or open	Default	Default mode: When INx is open, the corresponding channel output goes to its default logic state. Default is <i>High</i> for ISO7731B and <i>Low</i> for ISO7731B with F suffix.
X	PU	X	L	Z	A low value of Output Enable causes the outputs to be high-impedance
PD	PU	X	H or open	Default	Default mode: When $V_{CCI}$ is unpowered, a channel output assumes the logic state based on the selected default option. Default is <i>High</i> for ISO7731B and <i>Low</i> for ISO7731B with F suffix. When $V_{CCI}$ transitions from unpowered to powered-up, a channel output assumes the logic state of its input. When $V_{CCI}$ transitions from powered-up to unpowered, channel output assumes the selected default state.
X	PD	X	X	Undetermined	When $V_{CCO}$ is unpowered, a channel output is undetermined <sup>(3)</sup> . When $V_{CCO}$ transitions from unpowered to powered-up, a channel output assumes the logic state of its input

- (1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ ; PU = Powered up ( $V_{CC} \geq 2.25$  V); PD = Powered down ( $V_{CC} \leq 1.7$  V); X = Irrelevant; H = High level; L = Low level ; Z = High Impedance  
 (2) A strongly driven input signal can weakly power the floating  $V_{CC}$  through an internal protection diode and cause undetermined output.  
 (3) The outputs are in undetermined state when  $1.7$  V <  $V_{CCI}$ ,  $V_{CCO}$  < 2.25 V.

8.4.1 Device I/O Schematics



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Figure 15. Device I/O Schematics

## 9 Application and Implementation

### NOTE

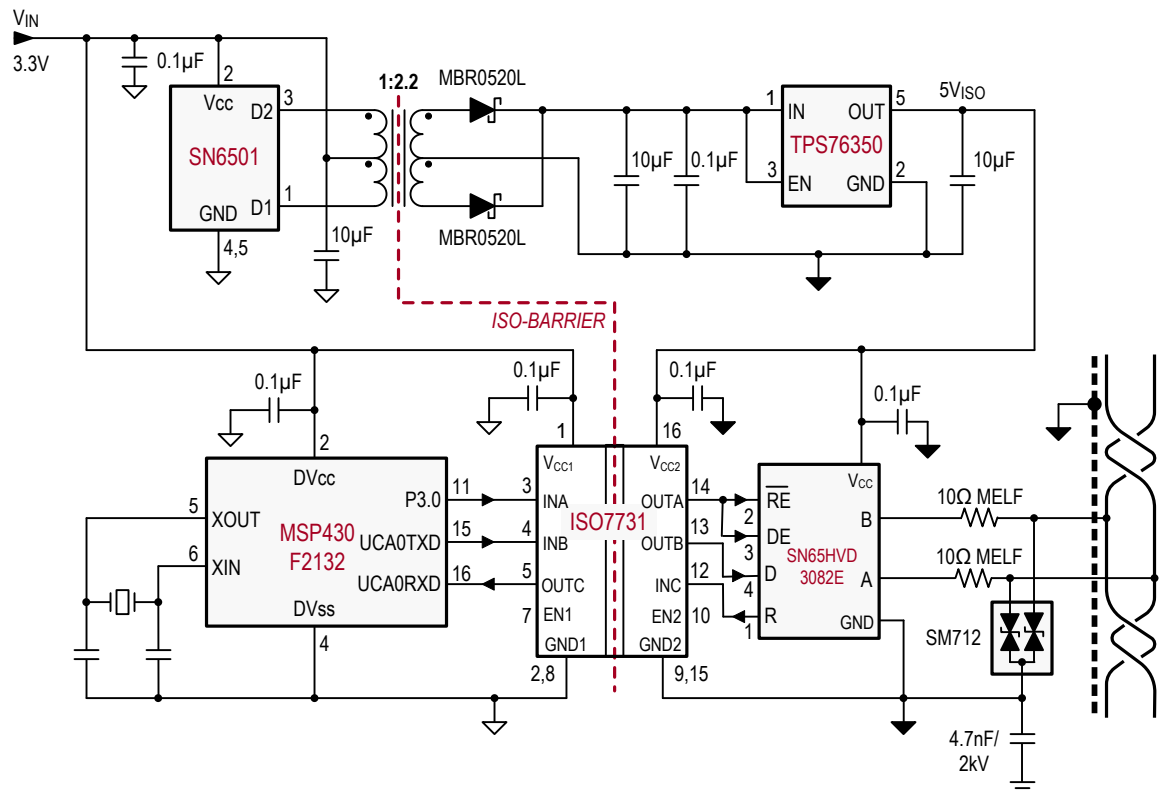
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

The ISO7731B device is a high-performance, triple-channel digital isolator. This device comes with enable pins on each side which can be used to put the respective outputs in high impedance for multi-master driving applications and reduce power consumption. The ISO7731B device uses single-ended CMOS-logic switching technology. The voltage range is from 2.25 V to 5.5 V for both supplies,  $V_{CC1}$  and  $V_{CC2}$ . When designing with digital isolators, keep in mind that because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is,  $\mu\text{C}$  or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

### 9.2 Typical Application

The ISO7731B device, combined with Texas Instruments' mixed-signal microcontroller, RS-485 transceiver, transformer driver, and voltage regulator, can create an isolated RS-485 system as shown in [Figure 16](#).



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**Figure 16. Isolated RS-485 Interface Circuit**

## Typical Application (continued)

### 9.2.1 Design Requirements

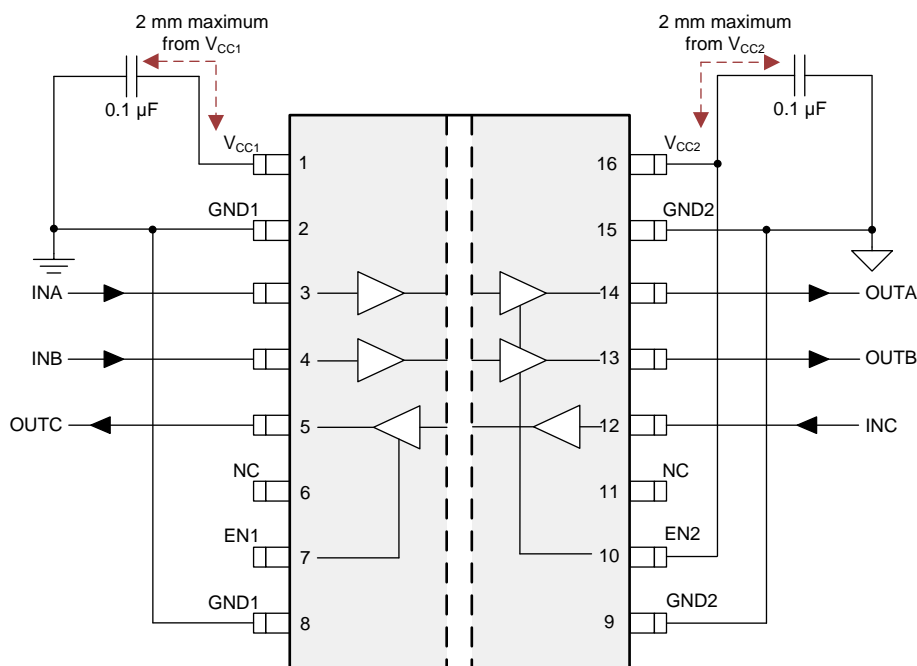
To design with these devices, use the parameters listed in [Table 3](#).

**Table 3. Design Parameters**

PARAMETER	VALUE
Supply voltage, $V_{CC1}$ and $V_{CC2}$	2.25 to 5.5 V
Decoupling capacitor between $V_{CC1}$ and GND1	0.1 $\mu$ F
Decoupling capacitor from $V_{CC2}$ and GND2	0.1 $\mu$ F

### 9.2.2 Detailed Design Procedure

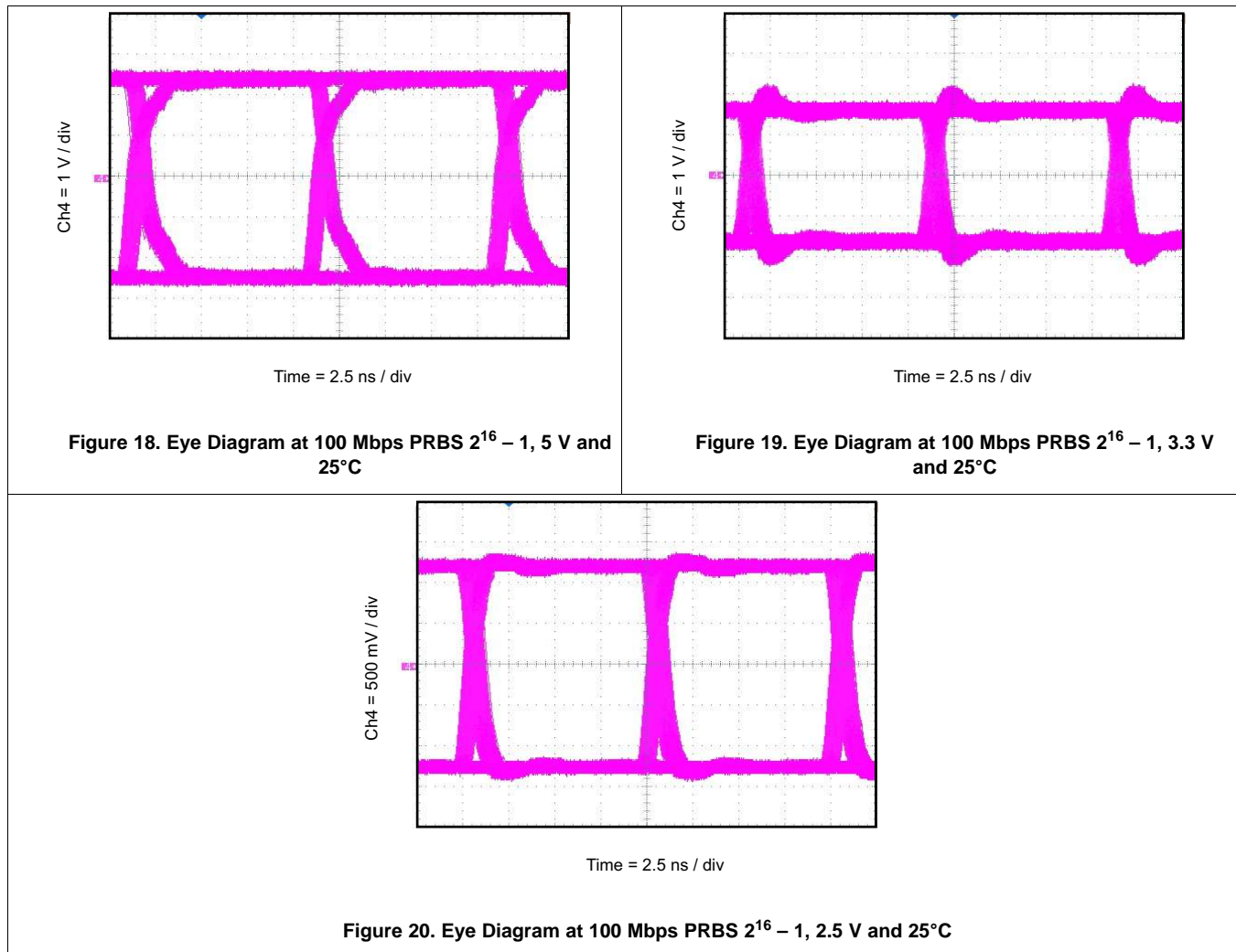
Unlike optocouplers, which require external components to improve performance, provide bias, or limit current, the ISO7731B device only requires two external bypass capacitors to operate. [Figure 17](#) shows the typical circuit hook-up for the device.



**Figure 17. Typical ISO7731B Circuit Hook-Up**

### 9.2.3 Application Curves

The following typical eye diagrams of the ISO7731B device indicates low jitter and wide open eye at the maximum data rate of 100 Mbps.



## 10 Power Supply Recommendations

To help ensure reliable operation at data rates and supply voltages, a 0.1- $\mu$ F bypass capacitor is recommended at the input and output supply pins ( $V_{CC1}$  and  $V_{CC2}$ ). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments' [SN6501](#) or [SN6505A](#). For such applications, detailed power supply design and transformer selection recommendations are available in the [SN6501 Transformer Driver for Isolated Power Supplies data sheet](#) or [SN6505A Low-Noise 1-A Transformer Drivers for Isolated Power Supplies data sheet](#).

## 11 Layout

### 11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see [Figure 21](#)). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/inch<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, refer to the [Digital Isolator Design Guide](#).

#### 11.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit boards. This PCB is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and self-extinguishing flammability-characteristics.

### 11.2 Layout Example

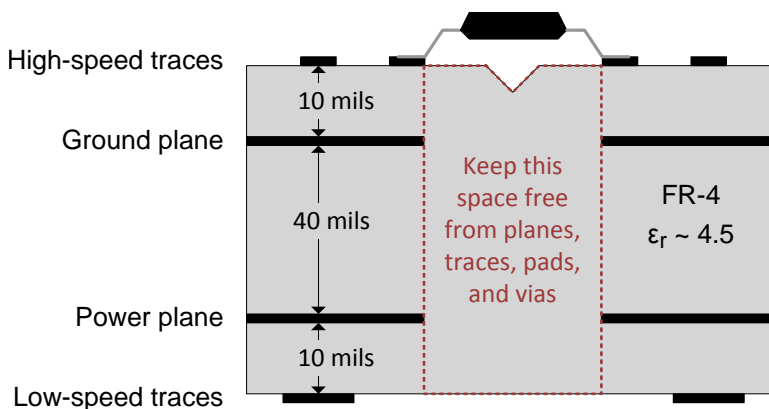


Figure 21. Layout Example Schematic

## 12 Device and Documentation Support

### 12.1 Documentation Support

#### 12.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [Digital Isolator Design Guide](#)
- Texas Instruments, [Isolation Glossary](#)
- Texas Instruments, [How to use isolation to improve ESD, EFT and Surge immunity in industrial systems application report](#)
- Texas Instruments, [SN6501 Transformer Driver for Isolated Power Supplies](#)
- Texas Instruments, [SNx5HVD308xE Low-Power RS-485 Transceivers, Available in a Small MSOP-8 Package](#)
- Texas Instruments, [TPS76350 Low-Power 150-mA Low-Dropout Linear Regulators](#)
- Texas Instruments, [MSP430F2132 Mixed Signal Microcontroller](#)

#### 12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 12.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 12.4 Trademarks

E2E is a trademark of Texas Instruments.

#### 12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 12.6 Glossary

**SLYZ022** — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

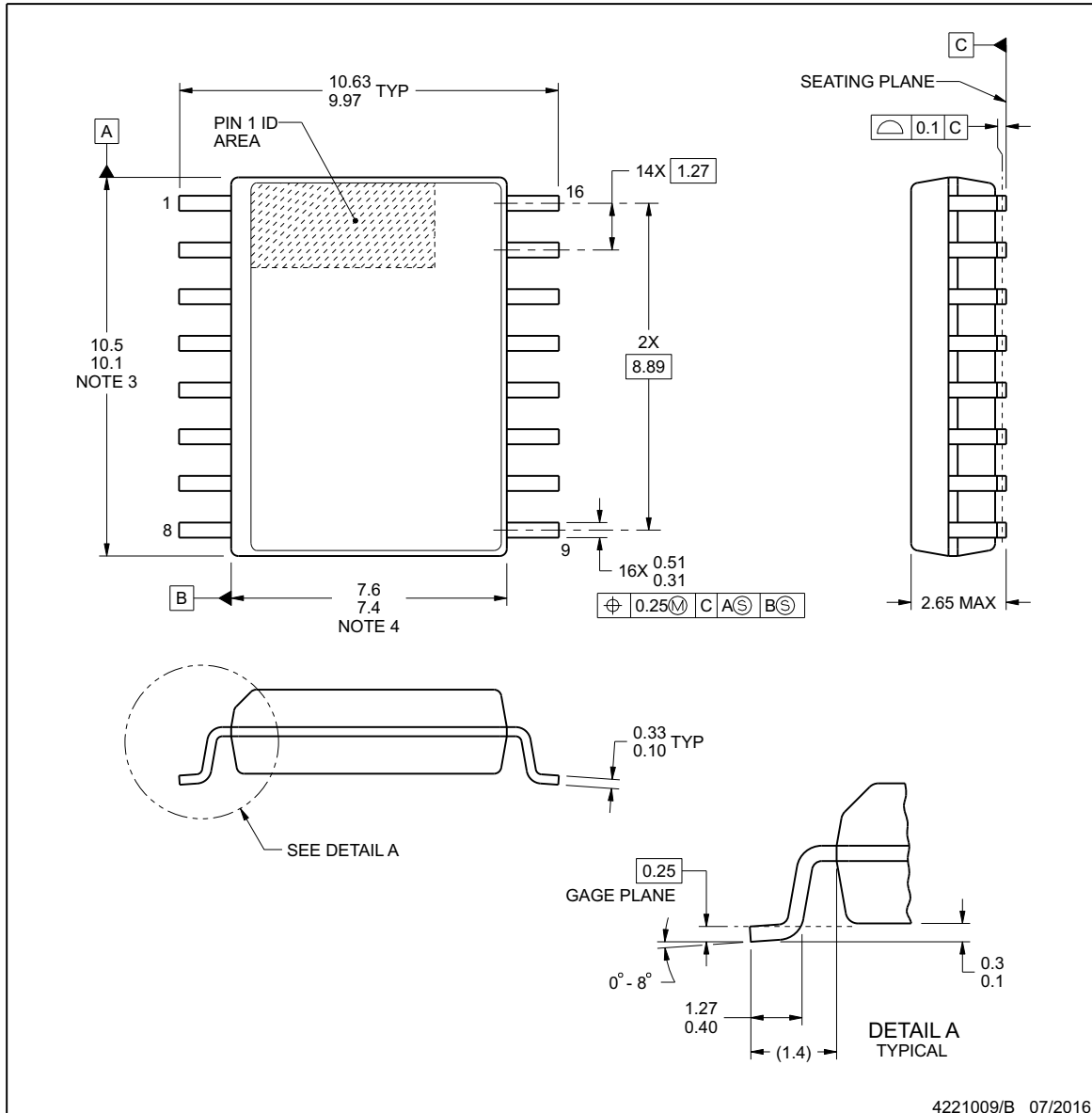


### **13 Mechanical, Packaging, and Orderable Information**

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.


**DW0016B**
**PACKAGE OUTLINE**
**SOIC - 2.65 mm max height**

SOIC


**NOTES:**

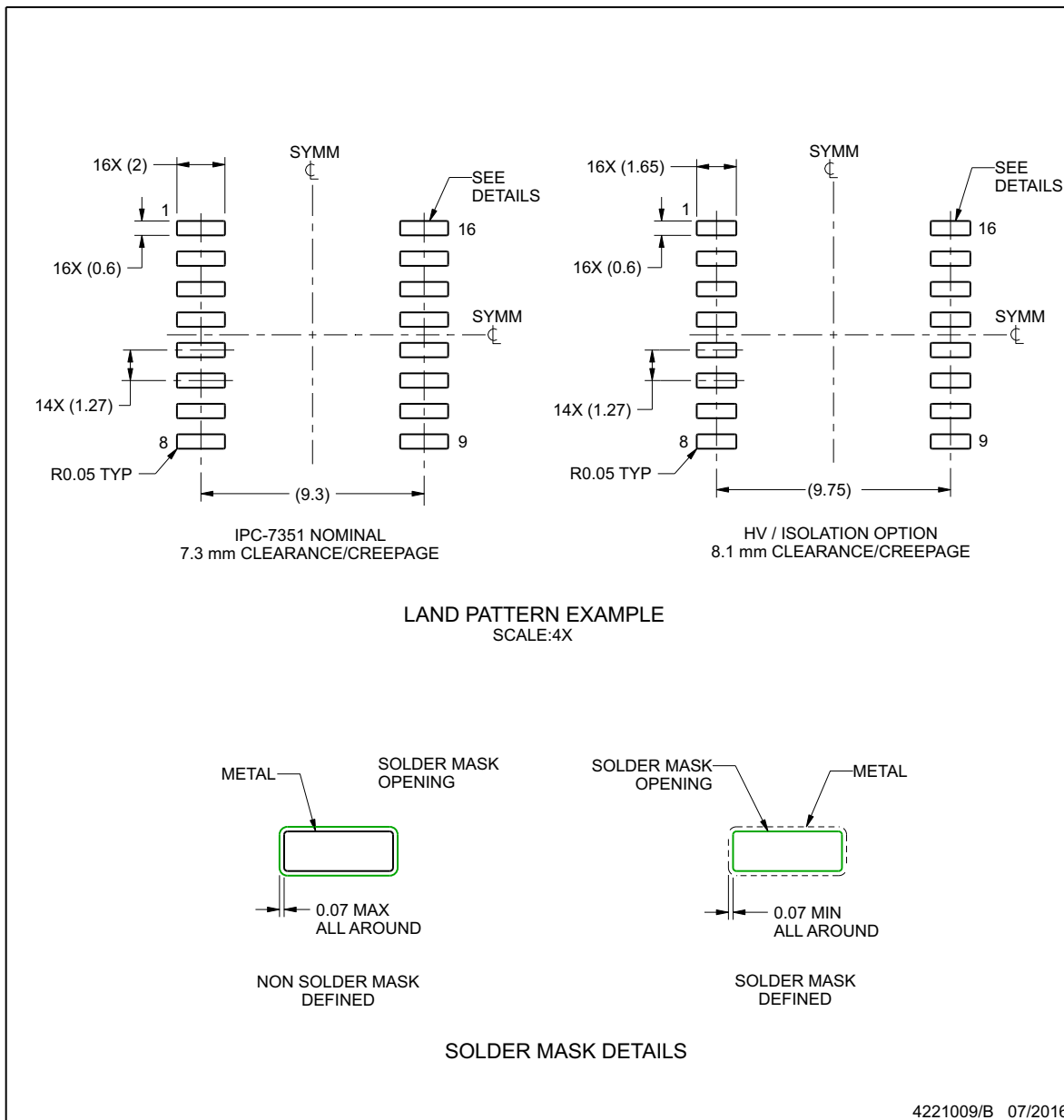
1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
5. Reference JEDEC registration MS-013.

## EXAMPLE BOARD LAYOUT

DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

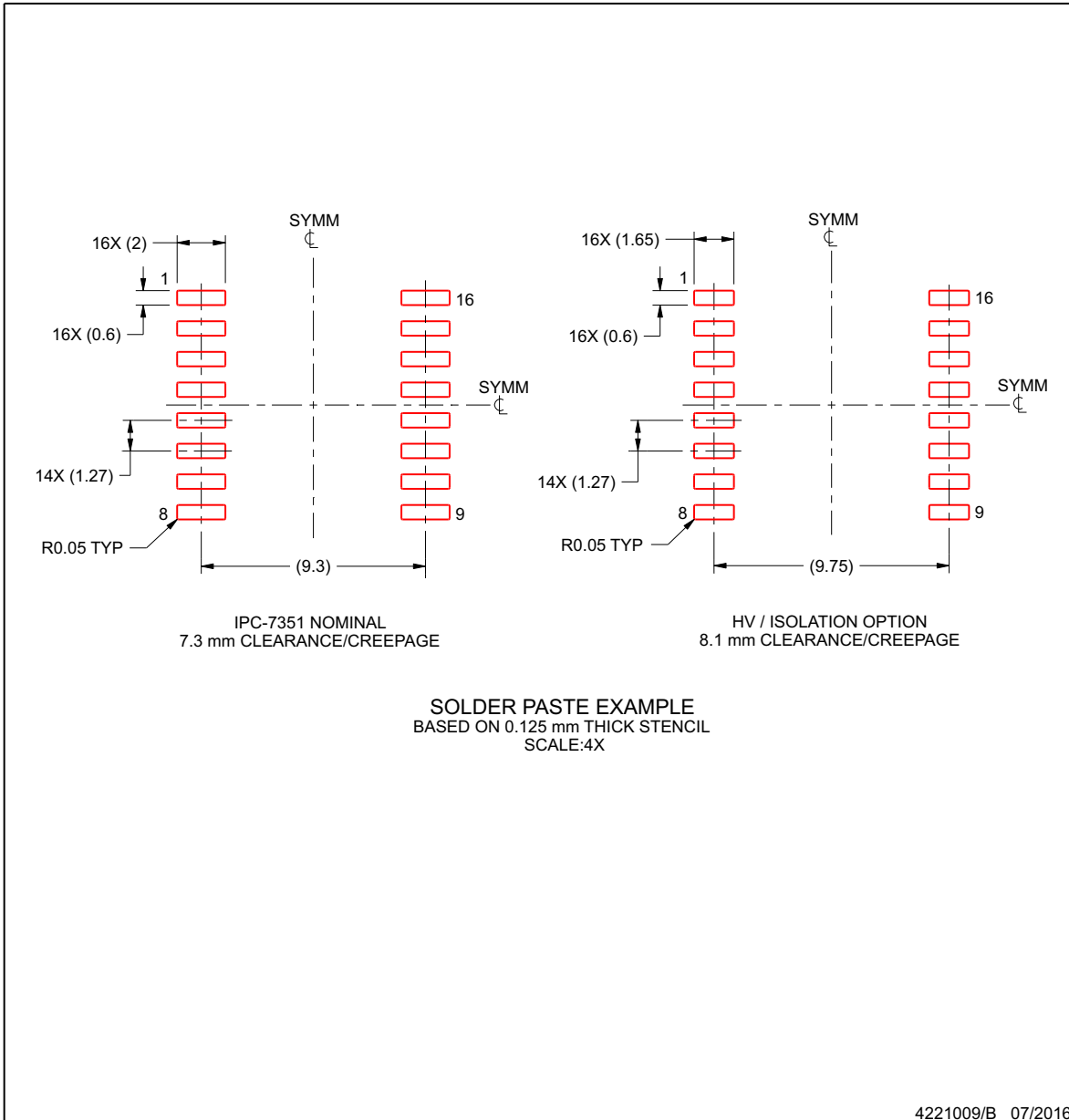
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

**EXAMPLE STENCIL DESIGN**

**DW0016B**

**SOIC - 2.65 mm max height**

SOIC



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.

## 13.1 Package Option Addendum

### 13.1.1 Packaging Information

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish <sup>(3)</sup>	MSL Peak Temp <sup>(4)</sup>	Op Temp (°C)	Device Marking <sup>(5)(6)</sup>
ISO7731BDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7731B
ISO7731BDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7731B
ISO7731FBDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7731FB
ISO7731FBDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7731FB

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PRE\_PROD** Unannounced device, not in production, not available for mass market, nor on the web, samples not available.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

(4) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(5) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device

(6) Multiple Device markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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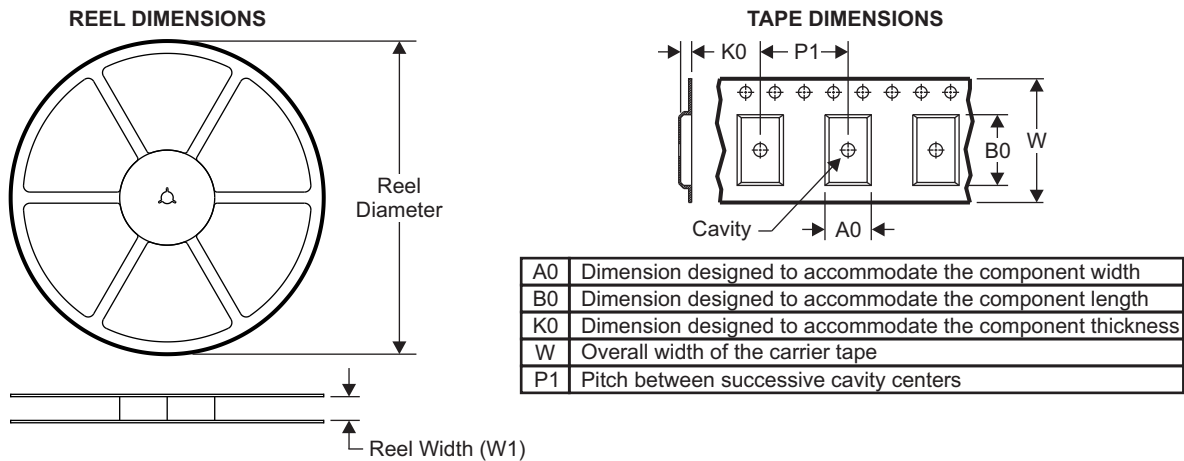
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# ISO7731B

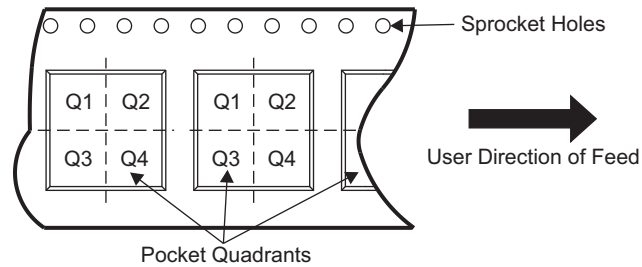
SLLSF65A – APRIL 2018 – REVISED JANUARY 2019

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## 13.1.2 Tape and Reel Information

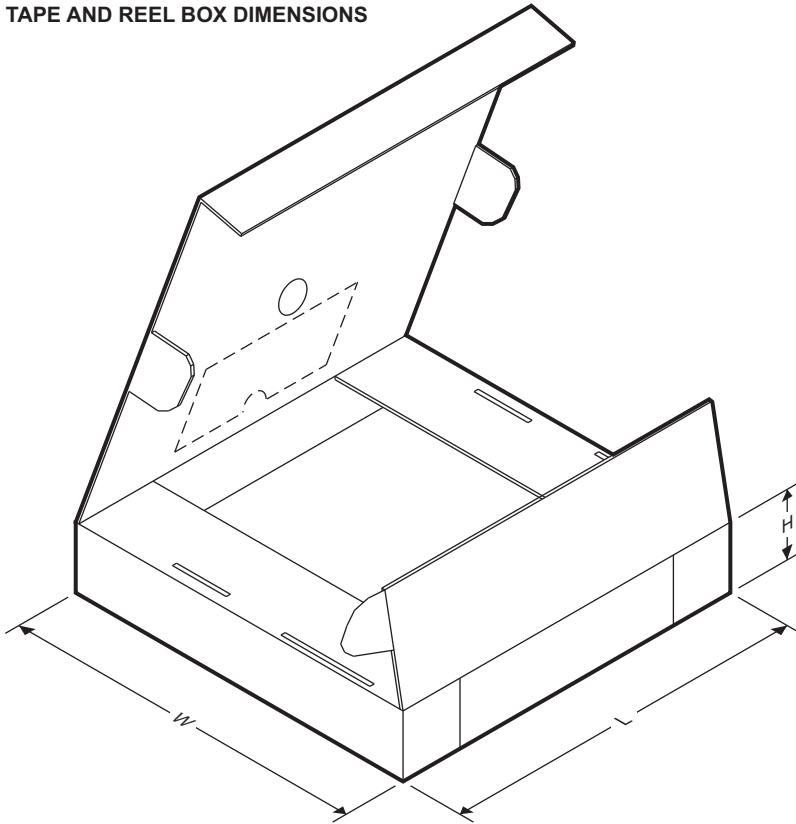


### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO7731BDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7731FBDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

**TAPE AND REEL BOX DIMENSIONS**



Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO7731BDWR	SOIC	DW	16	2000	367.0	367.0	38.0
ISO7731FBDWR	SOIC	DW	16	2000	367.0	367.0	38.0

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ISO7731BDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7731B	<a href="#">Samples</a>
ISO7731BDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7731B	<a href="#">Samples</a>
ISO7731FBDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7731FB	<a href="#">Samples</a>
ISO7731FBDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7731FB	<a href="#">Samples</a>

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(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

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## GENERIC PACKAGE VIEW

**DW 16**

**SOIC - 2.65 mm max height**

7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.



4224780/A



# DW0016B

# PACKAGE OUTLINE

## SOIC - 2.65 mm max height

SOIC



### NOTES:

1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
5. Reference JEDEC registration MS-013.

# EXAMPLE BOARD LAYOUT

DW0016B

SOIC - 2.65 mm max height

SOIC



LAND PATTERN EXAMPLE  
SCALE:4X



SOLDER MASK DETAILS

4221009/B 07/2016

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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